

Sensor and Simulation Notes

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## Focal Waveform of a Prolate-Spheroidal IRA

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### Abstract

This paper develops some analytic approximations for the focal waveform produced at the second focus of a prolate-spheroidal reflector due to a TEM wave launched from the first focus. This is extended to consider the spot size of the peak field near the second focus.

## 1. Introduction

Two previous papers [9, 10] have discussed the basic concept of using one or more prolate spheroidal reflectors to focus a fast-transient electromagnetic pulse on some targets of interest. Another paper [11] has shown that an inhomogeneous spherical TEM wave launched on guiding conical conductors from one focus is converted by a double stereographic transformation to a second (reflected) inhomogeneous spherical TEM wave propagating toward the second focus. Both waves have the same temporal waveforms before other scattering (from feed arms, etc.) can reach the observer.

The present paper is concerned with analytic calculations of the waveform at the second focus. For this purpose let us consider the geometry in Fig. 1.1. Let there be two thin perfectly conducting cone wave launchers with electrical centers lying in the  $xz$  plane. With respect to the negative  $z$  axis they are oriented at

$$\theta_1 = \theta_c \tag{1.1}$$

in the wave-launching spherical system  $(r_1, \theta_1, \phi_1)$ . These are related to cylindrical  $(\Psi_1, \phi_1, z_1)$  and Cartesian  $(x, y, z)$  coordinates as

$$\begin{aligned} \Psi_1 &= r_1 \sin(\theta_1) \quad , \quad z = -z_0 - r_1 \cos(\theta_1) \\ x &= \Psi_1 \cos(\phi_1) \quad , \quad y = -\Psi_1 \sin(\phi_1) \\ \phi_1 &= -\phi \quad , \quad \Psi = \Psi_1 \\ z_1 &= -z + z_0 \end{aligned} \tag{1.2}$$

The prolate spheroid is described by

$$\begin{aligned} \left[ \frac{\Psi}{b} \right]^2 + \left[ \frac{z}{a} \right]^2 &= 1 \\ z_0 &= \left[ a^2 - b^2 \right]^{1/2} \end{aligned} \tag{1.3}$$

The thin-cone electrical centers are described by the angle  $\theta_c$  with

$$\begin{aligned} \Psi_c &= r_1 \sin(\theta_c) \\ z_c &= -z_0 - r_1 \cos(\theta_c) \end{aligned} \tag{1.4}$$

At the reflector we have (subscript  $p$ )

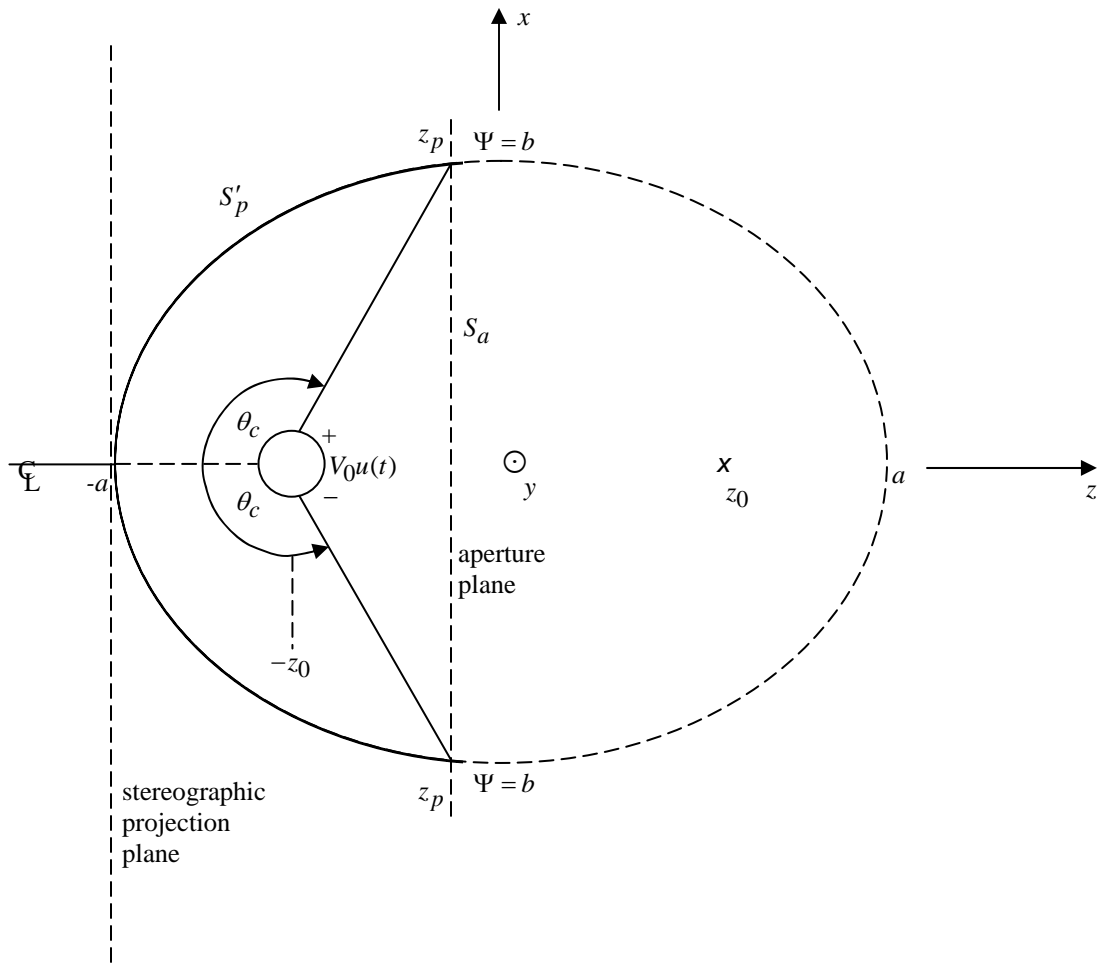


Fig. 1.1 Wave Launcher in Prolate-Spheroidal Reflector

$$\begin{aligned}
z_p &= -z_0 - \Psi_p \cot(\theta_c) \\
\left[ \frac{-z_0 - z_p}{b} \tan(\theta_c) \right]^2 + \left[ \frac{z_p}{a} \right]^2 &= 1 \\
\tan^2(\theta_c) &= \left[ \frac{b}{z_0 + z_p} \right]^2 \left[ 1 - \left[ \frac{z_p}{a} \right]^2 \right]
\end{aligned} \tag{1.5}$$

Given  $\theta_c$  one can solve for  $z_p$  and  $\Psi_p$ . One can also specify  $z_p$  and compute  $\Psi_p$  and  $\theta_c$ .

As a special case one may choose the symmetry plane for which

$$z_p = z = 0 \tag{1.6}$$

This simplifies the equations to

$$\begin{aligned}
\Psi_p &= b \quad , \quad r_p = a \\
\sin(\theta_c) &= \sin(\pi - \theta_c) = \frac{b}{a} \\
\frac{\pi}{2} &< \theta_c < \pi
\end{aligned} \tag{1.7}$$

For the present let us truncate the reflector at the  $z = z_p$  plane. The portion used is  $S'_p$ , to the left. This is consistent with the traditional truncation of paraboloidal reflector impulse radiating antennas (IRAs). More sophisticated truncation contours can be considered, but are beyond the scope of this paper. For later use the truncation plane will be taken as an aperture plane. The portion of this plane inside the prolate sphere is designated  $S_a$ . It is this surface which will be used for integrating over the reflected TEM wave to find the fields at the second focus,  $\vec{r}_0$ .

2. Prepulse

The stereographic projections in [11] can be used to calculate the fields. Let wave 1 have the form

$$\begin{aligned}
 \vec{E}_1 &= -r_1^{-1} \nabla_{\theta_1, \phi_1} V_1(\theta_1, \phi_1) u\left(t - \frac{r_1}{c}\right) \\
 \nabla_{\theta_1, \phi_1} &= \vec{1}_{\theta_1} \frac{\partial}{\partial \theta_1} + \vec{1}_{\phi_1} \frac{1}{\sin(\theta_1)} \frac{\partial}{\partial \phi_1} \\
 \nabla_{\theta_1, \phi_1}^2 V_1(\theta_1, \phi_1) &= 0 \\
 \nabla_{\theta_1, \phi_1}^2 &= \frac{1}{\sin(\theta_1)} \frac{\partial}{\partial \theta_1} \sin(\theta_1) \frac{\partial}{\partial \theta_1} + \frac{1}{\sin^2(\theta_1)} \frac{\partial^2}{\partial \phi^2} \\
 c &= [\mu \varepsilon]^{-1/2} = \text{wave speed in medium}
 \end{aligned} \tag{2.1}$$

Let the wave have  $V_1 = \pm V_0$  on the two cones.

The stereographic projection of this wave is

$$\Psi_0 = 2[a - z_0] \tan\left(\frac{\theta_1}{2}\right), \quad \phi_0 = -\phi_1 = -\phi \tag{2.2}$$

The electrical center of the thin wire on this projection plane is

$$\begin{aligned}
 \Psi_{c0} &= 2[a - z_0] \tan\left(\frac{\theta_c}{2}\right), \quad \phi_{c0} = 0, \pi \\
 r_{w0} &= \text{radius of thin wire on projection plane}
 \end{aligned} \tag{2.3}$$

For the two thin wires we have the well-known solution [1, 5]

$$\begin{aligned}
 V(x_0, y_0) &= V_0 \operatorname{arccosh}^{-1}\left(\frac{\Psi_{c0}}{r_{w0}}\right) \frac{1}{2} \ell_n \left( \frac{\left[ \frac{y_0}{\Psi_{c0}} \right]^2 + \left[ 1 + \frac{x_0}{\Psi_{c0}} \right]^2}{\left[ \frac{y_0}{\Psi_{c0}} \right]^2 + \left[ 1 - \frac{x_0}{\Psi_{c0}} \right]^2} \right) \\
 \operatorname{arccosh}\left(\frac{\Psi_{c0}}{r_{w0}}\right) &\approx \ell_n\left(2 \frac{\Psi_{c0}}{r_{w0}}\right)
 \end{aligned} \tag{2.4}$$

From this we have

$Z_c = Z_0 f_g \equiv$  characteristic impedance

$$Z_0 = \left[ \frac{\mu}{\varepsilon} \right]^{1/2} \equiv \text{wave impedance of medium} \quad (2.5)$$

$$f_g = \frac{1}{\pi} \operatorname{arccosh} \left( \frac{\Psi_{c0}}{r_{w0}} \right) \approx \frac{1}{\pi} \ell n \left( 2 \frac{\Psi_{c0}}{r_{w0}} \right)$$

It is often convenient to specify  $Z_c$  (around  $400 \Omega$ ), and from this determine the various transmission-line parameters. In this case we have

$$f_g \approx \frac{400}{377} \approx 1.06, \quad \frac{\Psi_{c0}}{r_{w0}} \approx 14 \quad (2.6)$$

for typical numbers.

Converting (2.4) to  $(\theta_1, \phi_1)$  coordinates we have

$$V_1(\theta_1, \phi_1) = \frac{V_0}{2} \operatorname{arccosh}^{-1} \left( \frac{\Psi_{c0}}{r_{w0}} \right) \ell n \left( \frac{\left[ \frac{\Psi_{c0}}{r_{w0}} \right]^2 + 2 \frac{\Psi_0}{\Psi_{c0}} \cos(\phi_1) + 1}{\left[ \frac{\Psi_0}{\Psi_{c0}} \right]^2 - 2 \frac{\Psi_0}{\Psi_{c0}} \cos(\phi_1) + 1} \right) \quad (2.7)$$

$$\Psi_0 = 2[a - z_0] \tan \left( \frac{\theta_1}{2} \right)$$

for the potential in wave 1. The gradient in (2.1) then gives the electric field. A special case has

$$\begin{aligned} \phi_1 &= \pm \pi/2 \\ V_1 &= 0 \end{aligned} \quad (2.8)$$

which is a symmetry plane (the  $x = 0$  plane). Another symmetry plane (the  $y = 0$  plane) has

$$\begin{aligned} \phi_1 &= 0 \quad (\text{for } \Psi_0 = +x) \\ V_1(\theta_1, \phi_1) &= V_0 \operatorname{arccosh}^{-1} \left( \frac{\Psi_{c0}}{r_{w0}} \right) \ell n \left( \frac{\frac{\Psi_0}{\Psi_{c0}} + 1}{\frac{\Psi_0}{\Psi_{c0}} - 1} \right) \end{aligned}$$

$$\begin{aligned}
&= V_0 \operatorname{arccosh}^{-1} \left( \frac{\Psi_{c0}}{r_{w0}} \right) \ln \left( \frac{1 + \frac{\Psi_{c0}}{\Psi_0}}{1 - \frac{\Psi_{c0}}{\Psi_0}} \right) = V_0 \operatorname{arccosh}^{-1} \left( \frac{\Psi_{c0}}{r_{w0}} \right) \left[ 2 \frac{\Psi_{c0}}{\Psi_0} + \mathcal{O} \left( \left[ \frac{\Psi_{c0}}{\Psi_0} \right]^2 \right) \right] \\
&= V_0 \operatorname{arccosh}^{-1} \left( \frac{\Psi_{c0}}{r_{w0}} \right) \left[ \frac{\Psi_{c0}}{a - z_0} \cot \left( \frac{\theta_1}{2} \right) + \dots \right]
\end{aligned} \tag{2.9}$$

For  $\theta_1$  near  $\pi$  this is

$$\begin{aligned}
V_1(\theta_1, \phi_1) &= V_0 \operatorname{arccosh}^{-1} \left( \frac{\Psi_{c0}}{r_{w0}} \right) \left[ \frac{\Psi_{c0}}{a - z_0} \frac{\pi - \theta_1}{2} + \dots \right], \quad \Psi_{c0} = 2[a - z_0] \tan \left( \frac{\theta_c}{2} \right) \\
V_1(\theta_1, \phi_1) &= V_0 \operatorname{arccosh}^{-1} \left( \frac{\Psi_{c0}}{r_{w0}} \right) \left[ \tan \left( \frac{\theta_c}{2} \right) [\pi - \theta_1] + \dots \right], \quad \pi f_g = \operatorname{arccosh} \left( \frac{\Psi_{c0}}{r_{w0}} \right) \approx \ln \left( 2 \frac{\Psi_{c0}}{r_{w0}} \right)
\end{aligned} \tag{2.10}$$

From (2.1) we have at the second focus

$$\begin{aligned}
\eta_1 &= 2z_0 \\
\vec{E}_1 &= -E_p u \left( t - \frac{2z_0}{c} \right) \vec{1}_x \\
E_p &= \frac{V_0}{4z_0} \operatorname{arccosh}^{-1} \left( \frac{\Psi_{c0}}{r_{w0}} \right) \frac{\Psi_{c0}}{a - z_0} \approx \frac{V_0}{2z_0 \pi f_g} \tan \left( \frac{\theta_c}{2} \right)
\end{aligned} \tag{2.11}$$

This is the prepulse orientation and magnitude (for step excitation). It lasts for a time [11]

$$\Delta t_p = \frac{2[a - z_0]}{c} = \frac{2}{c} \left[ a - [a^2 - b^2]^{1/2} \right] = \frac{2a}{c} \left[ 1 - \left[ 1 - \left[ \frac{b}{a} \right]^2 \right]^{1/2} \right] \tag{2.12}$$

after which the reflector signal arrives at the second focus.

For the special case as in (1.6) and (1.7) for which the launcher intersects the reflector at the  $z = 0$  symmetry plane we have

$$\begin{aligned}
\tan \left( \frac{\theta_c}{2} \right) &= \frac{1 - \cos(\theta_c)}{\sin(\theta_c)} = \frac{a}{b} \left[ 1 + \left[ 1 - \left[ \frac{b}{a} \right]^2 \right]^{1/2} \right] = \frac{1}{b} [a + z_0] \\
E_p &\approx \frac{V_0}{2\pi f_g} \frac{a + z_0}{b z_0} \\
E_p \Delta t_p &\approx \frac{V_0}{\pi f_g} \frac{b}{c z_0} \quad (\text{time integral or "area" of prepulse})
\end{aligned} \tag{2.13}$$

### 3. Fields on Aperture Plane

Now we consider the wave heading from the reflector toward the second focus. There is no set of conical conductors guiding the wave there. So we consider this second spherical TEM wave [11] on the aperture plane, which we can use in turn to find the fields at  $\vec{r}_0$  (and other positions as well). For present purposes we take the aperture plane as  $z = z_p$ , as in Fig. 1.1. The reflected wave illuminates  $S_a$ , a disk of radius  $\Psi_p$ .

Note that the reflector is truncated at the aperture plane. This is because the field from the wavelauncher reverses sign for the wave on the “other side” of the launching conductors. This is the same truncation used for typical reflector IRAs, based on a self-reciprocal geometry [12]. There are more sophisticated (nonplanar) truncation geometries which may be considered in the future.

In [11] the reflected wave was related to the first wave by a double stereographic transformation. They are equal (except for a minus sign) on the stereographic projection plane ( $z = -a$ ) for which

$$\begin{aligned}
 V_2(\Psi_0, \phi_0) &= -V_1(\Psi_0, \phi_0) \\
 (\Psi_0, \phi_0) &\equiv \text{cylindrical coordinates on projection plane} \\
 \Psi_0 &= 2[a - z_0] \tan\left(\frac{\theta_1}{2}\right) = 2[a + z_0] \tan\left(\frac{\theta_2}{2}\right) \\
 \phi_0 &= -\phi_1 = -\phi_2
 \end{aligned} \tag{3.1}$$

The second TEM wave takes the form

$$\vec{E}_2 = -\vec{r}_2^{-1} \nabla_{\theta_2, \phi_2} V_2(\theta_2, \phi_2) \mu \left( t + \frac{r_2}{c} - 2\frac{a}{c} \right) \tag{3.2}$$

Note the factor giving an incoming (on  $\vec{r}_0$ ) step-function wave.

For the aperture integral we need the tangential electric field on  $S_a$ . First we have

$$\begin{aligned}
 V_2(\theta_2, \phi_2) &= -\frac{V_0}{2} \operatorname{arccosh}^{-1} \left( \frac{\Psi_{c0}}{r_{w0}} \right) \ell_n \left( \frac{\left[ \frac{\Psi_0}{\Psi_{c0}} \right]^2 + 2 \frac{\Psi_0}{\Psi_{c0}} \cos(\phi_1) + 1}{\left[ \frac{\Psi_0}{\Psi_{c0}} \right]^2 - 2 \frac{\Psi_0}{\Psi_{c0}} \cos(\phi_1) + 1} \right) \\
 \Psi_0 &= 2[a + z_0] \tan\left(\frac{\theta_2}{2}\right), \quad \Psi_{c0} = 2[a - z_0] \tan\left(\frac{\theta_c}{2}\right)
 \end{aligned} \tag{3.3}$$



For  $\theta_2$  near zero we have

$$\begin{aligned}
V_2(\theta_2, \phi_2) &= -\frac{V_0}{2} \operatorname{arccosh}^{-1}\left(\frac{\Psi_{c0}}{r_{w0}}\right) \left[ 4 \frac{\Psi_0}{\Psi_{c0}} \cos(\phi_1) + \dots \right] \\
&= -V_0 \operatorname{arccosh}^{-1}\left(\frac{\Psi_{c0}}{r_{w0}}\right) \left[ 2 \frac{a+z_0}{a-z_0} \cot\left(\frac{\theta_c}{2}\right) \frac{\theta_2}{2} \cos(\phi_1) + \dots \right] \\
&\quad \operatorname{arccosh}^{-1}\left(\frac{\Psi_{c0}}{r_{w0}}\right) \approx \pi f_g \\
\vec{E}_2 &= \frac{V_0}{r_2 \pi f_g} \frac{a+z_0}{a-z_0} \cot\left(\frac{\theta_c}{2}\right) u\left(t + \frac{r_2}{c} - \frac{2a}{c}\right) \vec{1}_x
\end{aligned} \tag{3.4}$$

Comparing this to the prepulse amplitude in (2.11) we have ( $\theta_2$  near 0)

$$\vec{E}_2 = E_p \frac{2z_0}{r_2} \frac{a+z_0}{a-z_0} \cot^2\left(\frac{\theta_c}{2}\right) \vec{1}_x u\left(t + \frac{r_2}{c} - \frac{2a}{c}\right) \tag{3.5}$$

For the special case (as in (2.13)) that the launcher intersects the reflector at the  $z = 0$  symmetry plane we have (for  $\theta_2$  near zero)

$$\begin{aligned}
\vec{E}_2 &= \frac{V_0}{\pi f_g} \frac{z_0}{r_2} \frac{b}{a-z_0} u\left(t + \frac{r_2}{c} - \frac{2a}{c}\right) \vec{1}_x = E_p \frac{2z_0}{r_2} \frac{b^2}{[a-z_0][a+z_0]} u\left(t + \frac{r_2}{c} - \frac{2a}{c}\right) \vec{1}_x \\
&= E_p \frac{2z_0}{r_2} u\left(t + \frac{r_2}{c} - \frac{2a}{c}\right)
\end{aligned} \tag{3.6}$$

a fairly simple result.

At the center of the aperture plane we have

$$\begin{aligned}
r_2 &= z_0 - z_p, \quad \vec{E}_2 = E_{a0} u\left(t + \frac{z_0 - z_p}{c} - \frac{2a}{c}\right) \vec{1}_x \\
E_{a0} &= \frac{V_0}{\pi f_g} \frac{1}{z_0 - z_p} \frac{a+z_0}{a-z_0} \cot\left(\frac{\theta_c}{2}\right) \\
&= E_p \frac{2z_0}{z_0 - z_p} \frac{a+z_0}{a-z_0} \cot^2\left(\frac{\theta_c}{2}\right)
\end{aligned} \tag{3.7}$$

For  $z_p = 0$  this reduces to

$$E_{a0} = 2 E_p \tag{3.8}$$

Wave number 2 is focused on  $\vec{r}_0$ . Without guiding conductors (3.5) cannot hold all the way as  $r_2 \rightarrow 0$ . So we are considering the fields on  $S_a$  for later integration. On the center of  $S_a$  we have the electric field  $E_0$  polarized in the  $\vec{1}_x$  direction.

In IRA-related calculations [5, 6] it has been seen that, for circular apertures, the field at the center is an important parameter. The boresight radiated field can be found by integrating the TEM field over the aperture, or by integrating a uniform field of the center value (including polarization) over the same circular aperture. Seen another way, one can expand the field in cylindrical coordinates and note that terms with  $\cos(m\phi)$  and  $\sin(m\phi)$  for  $m \geq 2$  integrate to zero (for observation field points on the  $z$  axis). (There is no  $m=0$  term.) This is basically a symmetry result.

Similarly here, let us consider a uniform field on the  $(\Psi_0, \phi_0)$  projection plane. This corresponds to a potential function

$$V'_2 = -E_0 x_0 = -E_0 \Psi_0 \cos(\phi_0) \quad (3.9)$$

for a uniform field  $E_0$  polarized in the  $x$  direction. Matching this field to the second wave at

$$\begin{aligned} r_2 &= a + z_0 \text{ (center of projection plane)} \\ \vec{E}_2 &= E_0 u \left( t - \frac{z_0 - a}{c} \right) \vec{1}_x \\ E_0 &\approx \frac{V_0}{\pi f_g} \frac{1}{a - z_0} \cot \left( \frac{\theta_c}{2} \right) \\ &= E_p \frac{2z_0}{a - z_0} \cot^2 \left( \frac{\theta_c}{2} \right) \end{aligned} \quad (3.10)$$

For  $z_p = 0$  this becomes

$$E_0 \approx E_p \frac{2z_0}{a + z_0} \approx \frac{V_0}{\pi f_g b} \quad (3.11)$$

Now convert  $V'_2$  on the projection plane to  $(\theta_2, \phi_2)$  coordinates as

$$V'_2 = -E_0 2[a + z_0] \tan \left( \frac{\theta_2}{2} \right) \cos(\phi_2) \quad (3.12)$$

giving an electric field

$$\begin{aligned}
\vec{E}_2 &= E_0 \frac{a+z_0}{r_2} \left[ \sec^2\left(\frac{\theta_2}{2}\right) \cos(\theta_2) \vec{1}_{\theta_2} - 2 \tan\left(\frac{\theta_2}{2}\right) \frac{\sin(\phi_2)}{\sin(\theta_2)} \vec{1}_{\phi_2} \right] u\left(t + \frac{r_2}{c} - 2\frac{a}{c}\right) \\
&= E_0 \frac{a+z_0}{r_2} \left[ \frac{2\cos(\phi_2)}{1+\cos(\theta_2)} \vec{1}_{\theta_2} - \frac{2\sin(\phi_2)}{1+\cos(\theta_2)} \vec{1}_{\phi_2} \right] u\left(t + \frac{r_2}{c} - 2\frac{a}{c}\right)
\end{aligned} \tag{3.13}$$

On the aperture plane we have

$$\begin{aligned}
r_2 &= \left[ \Psi^2 + [z_0 - z_p]^2 \right]^{1/2} \\
\cos(\theta_2) &= \frac{z_0 - z_p}{r_2} \\
r_2 [1 + \cos(\theta_2)] &= r_2 + z_0 - z_p \\
\phi &= -\phi_2, \quad \vec{1}_{\phi} = -\vec{1}_{\phi_2} \\
\vec{1}_{\theta_2} \cdot \vec{1}_{\Psi} &= \cos(\theta_2)
\end{aligned} \tag{3.14}$$

The tangential electric field is then

$$\begin{aligned}
\vec{E}_{2t} &= 2E_0 \frac{a+z_0}{r_2 + z_0 - z_p} \left[ \frac{z_0 - z_p}{r_2} \cos(\phi) \vec{1}_{\Psi} - \sin(\phi) \vec{1}_{\phi} \right] \\
&\quad u\left(t + \frac{r_2}{c} - 2\frac{a}{c}\right)
\end{aligned} \tag{3.15}$$

Converting to Cartesian coordinates we need only the  $x$  component (due to symmetry) as

$$\begin{aligned}
\vec{1}_{\Psi} \cdot \vec{1}_x &= \cos(\phi) = \vec{1}_{\phi} \cdot \vec{1}_y = -\sin(\phi) \\
E'_{2x} &= 2E_0 \frac{a+z_0}{r_2 + z_0 - z_p} \left[ \frac{z_0 - z_p}{r_2} \cos^2(\phi) + \sin^2(\phi) \right] \\
E_{2x} &= E'_{2x} u\left(t + \frac{r_2}{c} - 2\frac{a}{c}\right)
\end{aligned} \tag{3.16}$$

4. Fields at Second Focus

We are now in a position to evaluate the fields at  $\vec{r}_0$  by integrals over the fields on  $S_a$ . A previous paper [2] has developed the formulae. From [2 (3.3)] we have (in time domain for step excitation)

$$\begin{aligned}\vec{E}_f &= E_{fx} \vec{1}_x, \quad E_{fx} = E_\delta \delta\left(t - 2\frac{a}{c}\right) + E_s u\left(t - 2\frac{a}{c}\right) \\ E_\delta &= \frac{1}{2\pi c} \int_{S_a} \frac{z_0 - z_p}{r_2^e} E'_{2x} ds, \quad E_s = \frac{1}{2\pi} \int_{S_a} \frac{z_0 - z_p}{r_2^3} E'_{2x} ds \\ r_2 &= \left[ [z_0 - z_p]^2 + \Psi^2 \right]^{1/2}\end{aligned}\tag{4.1}$$

Using cylindrical coordinates on  $S_a$  we can evaluate the integrals. Considering first the  $\phi$  variable we have

$$\begin{aligned}E_\delta &= \frac{E_0}{\pi c} \int_0^\Psi \int_0^{2\pi} \frac{z_0 - z_p}{r_2^2} \frac{a + z_0}{r_2 + z_0 - z_p} \left[ \frac{z_0 - z_p}{r_2} \cos^2(\phi) + \sin^2(\phi) \right] \Psi d\phi d\Psi \\ &= \frac{E_0}{c} \int_0^\Psi \frac{z_0 - z_p}{r_2^2} \frac{a + z_0}{r_2 + z_0 - z_p} \left[ \frac{z_0 - z_p}{r_2} + 1 \right] \Psi d\Psi \\ &= \frac{E_0}{c} [z_0 - z_p] [a + z_0] \int_0^\Psi \frac{\Psi}{r_2^3} d\Psi \\ E_s &= \frac{E_0}{\pi} \int_0^\Psi \int_0^{2\pi} \frac{z_0 - z_p}{r_2^3} \frac{a + z_0}{r_2 + z_0 - z_p} \left[ \frac{z_0 - z_p}{r_2} \cos^2(\phi) + \sin^2(\phi) \right] \Psi d\phi d\Psi \\ &= E_0 \int_0^\Psi \frac{z_0 - z_p}{r_2^3} \frac{a + z_0}{r_2 + z_0 - z_p} \left[ \frac{z_0 - z_p}{r_2} + 1 \right] \Psi d\Psi \\ &= E_0 [z_0 - z_p] [a + z_0] \int_0^\Psi \frac{\Psi}{r_2^4} d\Psi\end{aligned}\tag{4.2}$$

Next we have

$$\begin{aligned}r_2 &= \left[ [z_0 - z_p]^2 + \Psi^2 \right]^{1/2}, \quad v \equiv \Psi^2 + [z_0 - z_p]^2, \quad dv = 2\Psi d\Psi \\ E_\delta &= \frac{E_0}{2c} [z_0 - z_p] [a + z_0] \int_{[z_0 - z_p]^2}^{\Psi^2 + [z_0 - z_p]^2} v^{-3/2} dv\end{aligned}$$

$$\begin{aligned}
&= \frac{E_0}{c} [z_0 - z_p] [a + z_0] \left[ [z_0 - z_p]^{-1} \left[ \Psi_p^2 + [z_0 - z_p]^2 \right]^{-1/2} \right] \\
&= \frac{E_0}{c} [a + z_0] \left[ 1 - \left[ 1 + \left[ \frac{\Psi_p}{[z_0 - z_p]} \right]^2 \right]^{-1/2} \right] \\
E_S &= \frac{E_0}{2} [z_0 - z_p] [a + z_0] \int_{[z_0 - z_p]^2}^{\Psi_p^2 + [z_0 - z_p]^2} v^{-2} dv \\
&= \frac{E_0}{2} [z_0 - z_p] [a + z_0] \left[ [z_0 - z_p]^{-2} - \left[ \Psi_p^2 + [z_0 - z_p]^2 \right]^{-1} \right] \\
&= \frac{E_0}{2} \frac{a + z_0}{z_0 - z_p} \left[ 1 - \left[ 1 + \left[ \frac{\Psi_p}{[z_0 - z_p]} \right]^2 \right]^{-1} \right] \\
&= \frac{E_0}{2} \frac{a + z_0}{z_0 - z_p} \left[ 1 + \left[ \frac{z_0 - z_p}{\Psi_p} \right]^2 \right]^{-1} \tag{4.3}
\end{aligned}$$

Note that  $E_S$  has units V/m, while  $E_\delta$  has units Vs/m corresponding to the time integral (or area) of the  $\delta$  function.

## 5. Some Design Considerations

Summarizing, we have

$$\begin{aligned}
E_\delta &= \frac{E_0}{c} [a + z_0] \left[ 1 - \left[ 1 + \left[ \frac{\Psi_p}{z_0 - z_p} \right]^2 \right]^{-1/2} \right] = \frac{V_0}{\pi f_g c} \frac{a + z_0}{a - z_0} \cot\left(\frac{\theta_c}{2}\right) \left[ 1 - \left[ 1 + \left[ \frac{\Psi_p}{z_0 - z_p} \right]^2 \right]^{-1/2} \right] \\
E_s &= \frac{E_0}{2} \frac{a + z_0}{z_0 - z_p} \left[ 1 + \left[ \frac{z_0 - z_p}{\Psi_p} \right]^2 \right]^{-1} = \frac{V_0}{2\pi f_g} \frac{1}{z_0 - z_p} \frac{a + z_0}{a - z_0} \cot\left(\frac{\theta_c}{2}\right) \left[ 1 + \left[ \frac{z_0 - z_p}{\Psi_p} \right]^2 \right]^{-1} \\
E_p &= \frac{V_0}{2\pi f_g z_0} \tan\left(\frac{\theta_c}{2}\right) \quad , \quad E_p \Delta t_p = \frac{V_0}{2\pi f_g c} \frac{a - z_0}{z_0} \tan\left(\frac{\theta_c}{2}\right) \\
E_0 &= \frac{V_0}{\pi f_g} \frac{1}{a - z_0} \cot\left(\frac{\theta_c}{2}\right)
\end{aligned} \tag{5.1}$$

One can compute  $\theta_c$  from  $z_p$  (or  $\Psi_p$ ) in (1.5). For the special symmetric case of  $z_p = 0$ , merely replace  $\Psi_p = b$ , giving

$$\begin{aligned}
E_\delta &= \frac{E_0}{c} \frac{b^2}{a} \approx \frac{V_0}{\pi f_g c} \frac{b}{a} \quad , \quad E_s = \frac{E_0}{2} \frac{a + z_0}{z_0} \frac{b^2}{a^2} \approx \frac{V_0}{2\pi f_g} \frac{a + z_0}{z_0} \frac{b}{a^2} \\
E_p &\approx \frac{V_0}{2\pi f_g} \frac{a + z_0}{bz_0} \quad , \quad E_p \Delta t_p \approx \frac{V_0}{\pi f_g c} \frac{b}{z_0} \\
E_0 &= \frac{V_0}{\pi f_g b}
\end{aligned} \tag{5.2}$$

A first observation concerns the common factor  $V_0/(\pi f_g)$ . For large fields one needs large voltage and low wave-launcher impedance. Note that an extra factor of  $\sqrt{2}$  increase in the fields is obtained by going to a 4-arm feed in the usual sense of a IRA with arms at  $45^\circ$  [4], and a little more can be achieved by arms at about  $60^\circ$  from the horizontal (the  $x = 0$  plane) as projected on a constant- $z$  plane [7, 8]. This is just another common factor with which we can deal separately.

The  $\delta$ -function part of the field does not have infinite amplitude when the incident step-like wave has a nonzero rise time. For simplicity let us imagine that the  $\delta$ -function is replaced by

$$\delta(t) \rightarrow \frac{1}{t_\delta} [u(t) - u(t - t_\delta)] \tag{5.3}$$

i.e., a gate function of width  $t_\delta$  and height  $t_\delta^{-1}$ . This makes the impulsive part of the field have amplitude  $E_\delta t_\delta^{-1}$ , showing the importance of a small pulse width. For convenience define

$$T \equiv \frac{\Delta t_p}{t_\delta} = \frac{2[a-z_0]}{ct_\delta} \quad (5.4)$$

We will want this to be large, which argues for large  $a$ .

Considering the simpler case  $z_p = 0$  we have

$$\frac{E_s}{E_p} \approx \frac{b^2}{a^2} < 1 \quad (5.5)$$

If we want the step from the reflector (the post pulse if you like) to cancel the prepulse step then one would like  $b$  to approach  $a$ , but this has other consequences. Next compare the main ( $\delta$ ) pulse amplitude to the prepulse amplitude giving

$$\frac{E_\delta}{t_\delta E_p} \approx 2 \frac{b^2 z_0}{act_\delta} [a+z_0]^{-1} = 2 \frac{b^2}{a} \frac{T}{ct_p} \frac{z_0}{a+z_0} = \frac{z_0}{a} T \quad (5.6)$$

The factor  $z_0/a$  is less than one, but it needs to be large to give a large main pulse. This is further aided by a large  $T$  implying large  $a-z_0$  and small  $t_\delta$ . So, intermediate values of  $z_0$  are called for, i.e., at the maximum of  $z_0[a-z_0]$  which is

$$\begin{aligned} z_0 &= \frac{a}{2}, \quad b = \frac{\sqrt{3}}{2} a \\ \theta_c &= \frac{3}{4} \pi = 135^\circ \\ \frac{b^2}{a^2} &= \frac{3}{4} \end{aligned} \quad (5.7)$$

## 6. Spot Size

Given that the impulse has some small width  $t_\delta$ , the maximum fields will exist in some small region around  $\vec{r}_0$ . So let us estimate a pulse width of this impulsive part for positions near  $\vec{r}_0$ . For the  $\delta$ -function pulse we have the wave from every position on  $S_a$  arriving at exactly the same time at  $\vec{r}_0$ . We can then estimate a pulse width near  $\vec{r}_0$  by the dispersion in the arrival times from all parts of  $S_a$  at the observation point.

For this purpose, consider the lengths of the ray paths in Fig. 6.1 indicating the maximum time differences from the edges and center of  $S_a$  to the observer. For an observer at  $z_0 + \Delta z$  on the z axis we have

$t_c \equiv$  arrival time from center

$$ct_c = 2a + \Delta z$$

$t_e \equiv$  arrival time from edge

$$\begin{aligned} ct_e &= \left[ [z_0 + z_p]^2 + \Psi_p^2 \right]^{1/2} + \left[ [z_0 - z_p + \Delta z]^2 + \Psi_p^2 \right]^{1/2} \\ &\approx \left[ [z_0 + z_p]^2 + \Psi_p^2 \right]^{1/2} + \left[ [z_0 - z_p]^2 + 2[z_0 - z_p]\Delta z + \Psi_p^2 \right]^{1/2} \\ &\approx \left[ [z_0 + z_p]^2 + \Psi_p^2 \right]^{1/2} + \left[ [z_0 - z_p]^2 + \Psi_p^2 \right]^{1/2} \left[ 1 + \frac{[z_0 - z_p]\Delta z}{[z_0 - z_p]^2 + \Psi_p^2} \right] \\ &= 2a + \frac{[z_0 - z_p]\Delta z}{\left[ [z_0 - z_p]^2 + \Psi_p^2 \right]^{1/2}} \end{aligned} \quad (6.1)$$

Then we have

$$\begin{aligned} t_z &\equiv \text{pulse width with respect to } z = |t_c - t_e| \\ ct_z &\approx \left| \frac{z_0 - z_p}{\left[ [z_0 - z_p]^2 + \Psi_p^2 \right]^{1/2}} - 1 \right| |\Delta z| \end{aligned} \quad (6.2)$$

For the special case of  $z_p = 0$  we have

$$ct_z \approx \left[ 1 - \frac{z_0}{a} \right] |\Delta z| \quad (6.3)$$

Large  $z_0/a$  (small  $b/a$ ) minimizes this.

For an observer radially displaced  $\Delta\Psi$  from  $\vec{r}_0$  we have

$t_1 \equiv$  arrival time from near edge



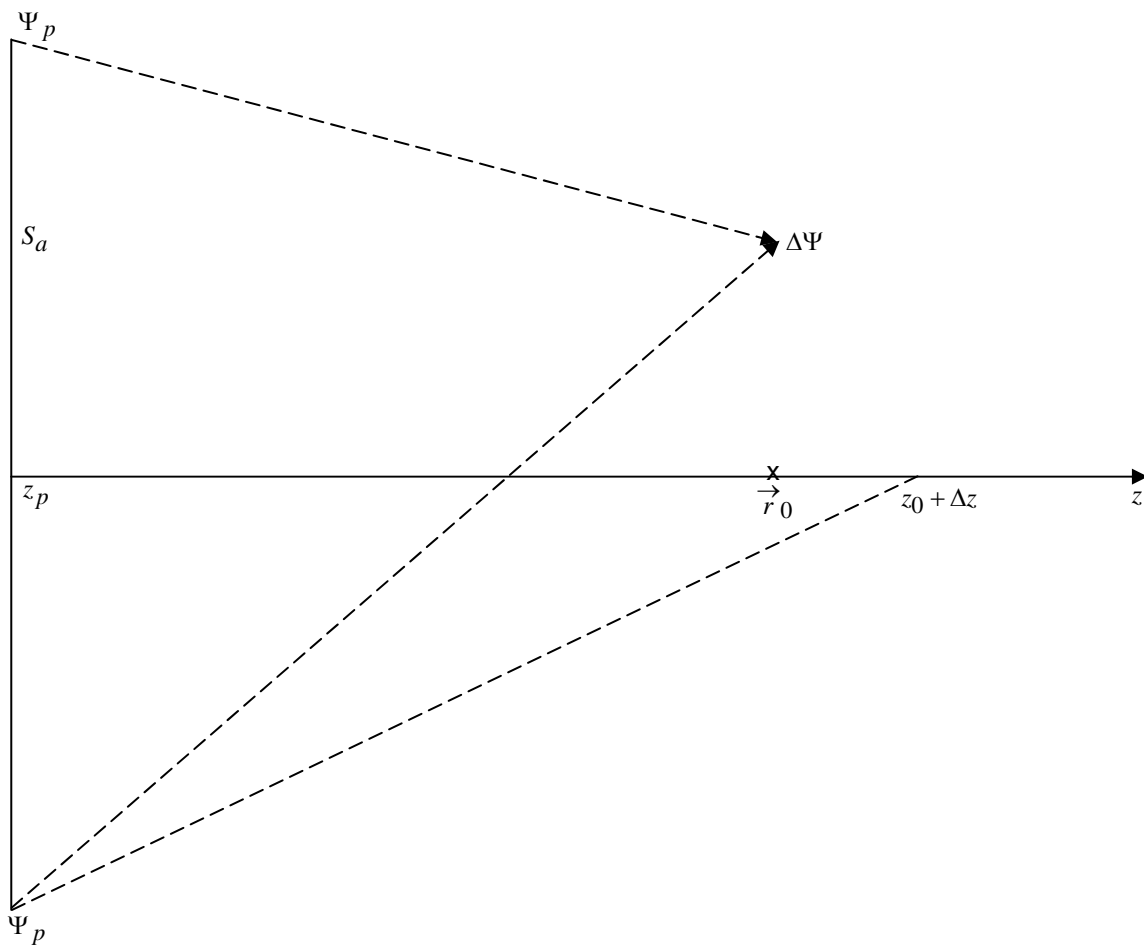


Fig. 6.1 Dispersion of Impulse Near  $\vec{r}_0$ .

$$\begin{aligned}
ct_1 &= \left[ [z_0 + z_p]^2 + \Psi_p^2 \right]^{1/2} + \left[ [z_0 - z_p]^2 + [\Psi_r - \Delta\Psi]^2 \right]^{1/2} \\
&\approx \left[ [z_0 + z_p]^2 + \Psi_p^2 \right]^{1/2} + \left[ [z_0 - z_p]^2 + \Psi_p^2 - 2\Psi_p\Delta\Psi \right]^{1/2} \\
&\approx \left[ [z_0 + z_p]^2 + \Psi_p^2 \right]^{1/2} + \left[ [z_0 - z_p]^2 + \Psi_p^2 \right]^{1/2} \left[ 1 - \frac{\Psi_r\Delta\Psi}{[z_0 - z_p]^2 + \Psi_p^2} \right] \\
&= 2a - \frac{\Psi_p\Delta\Psi}{\left[ [z_0 - z_p]^2 + \Psi_p^2 \right]^{1/2}}
\end{aligned}$$

$t_2 \equiv$  arrival time from far edge

$$ct_2 \approx 2a + \frac{\Psi_p\Delta\Psi}{\left[ [z_0 - z_p]^2 + \Psi_p^2 \right]^{1/2}} \quad (6.4)$$

Then we have

$t_\Psi \equiv$  pulse width with respect to  $\Psi = t_2 - t_1$

$$ct_\Psi = \frac{2\Psi_p\Delta\Psi}{\left[ [z_0 - z_p]^2 + \Psi_p^2 \right]^{1/2}} \quad (6.5)$$

For the special case of  $z_p = 0$  we have

$$ct_\Psi = 2\frac{b}{a}\Delta\Psi \quad (\Delta\Psi \text{ positive}) \quad (6.6)$$

Small  $b/a$  also minimizes this.

Comparing these pulse widths to  $t_\delta$ , the width at  $\vec{r}_0$ , we can note that the pulse widths in the  $z$  direction are  $t_\delta + t_z$ , and in the  $\Psi$  direction  $t_\delta + t_\Psi$ . The physical spot size is then given by (counting width in both directions from  $\vec{r}_0$ )

$$t_\Psi \approx t_z \approx 2t_\delta \quad (6.7)$$

Given by (6.2) and (6.4). For the special case of  $z_p = 0$  we have

$$|\Delta z| \approx 2\left[1 - \frac{z_0}{a}\right]^{-1} ct_\delta \quad , \quad \Delta\Psi \approx \frac{a}{b} ct_\delta \quad (6.8)$$

Of course, these are just rough estimates. Detailed waveform calculations will give more accurate results, including actual waveforms instead of bounds on pulse widths.

## 7. Concluding Remarks

This modest study has found some analytic approximations useful for estimating the focal waveforms and focal spot size for the prolate-spheroidal IRA. One may think of this as analogous to the first of the papers dealing with the IRA using a paraboloidal reflector [3]. Considering the sophisticated design papers which followed the introduction of the IRA concept, there is much yet to be done for the prolate-spheroidal version.

Here we have found some especially simple results for the case of reflector truncation at  $z_p = 0$ . More detailed numerical treatment of the results for  $z_p \neq 0$  may lead to yet more insight into an optimal design.

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