

Interaction Notes

Note 203

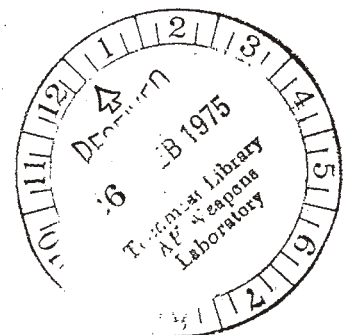
June 1974

Computer Program for Thin-Wire Structures
in a Homogeneous Conducting Medium

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ABSTRACT

A computer program is presented for thin-wire antennas and scatterers in a homogenous conducting medium. The analysis is performed in the real or complex frequency domain. The program handles insulated and bare wires with finite conductivity and lumped loads. The output data includes the current distribution, impedance, radiation efficiency, gain, absorption cross section, scattering cross section, echo area and the polarization scattering matrix. The program uses sinusoidal bases and Galerkin's method.



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I. INTRODUCTION

Reference 1 presents the electromagnetic theory for a thin-wire structure in a homogeneous conducting medium, and this report presents the corresponding computer program. The program performs a frequency-domain analysis of thin-wire antennas and scatterers. The wire configuration is a generalized polygon assembled from straight wire segments. The program has been tested extensively with simple structures (linear dipoles, V dipoles, coupled dipoles, square loops, octagonal loops, multiturn loops and coupled loops) and complicated configurations including wire-grid models of plates, spheres, cones, aircraft and ships. Although the air-earth or air-water interface is not considered, the program is applicable in many problems involving buried or submerged antennas or targets. It is useful in locating the poles of the admittance or scattering function for wire structures in the complex frequency domain.

A piecewise-sinusoidal expansion is used for the current distribution. The matrix equation $ZI = V$ is generated by enforcing reaction tests with a set of sinusoidal dipoles located in the interior region of the wire. Since the test dipoles have the same current distribution as the expansion modes, this may be regarded as an application of Galerkin's method. Rumsey's reaction concept was most helpful in this development, and therefore the formulation is known as the "sinusoidal reaction technique."

The current is assumed to vanish at the endpoints (if any) of the wire, and Kirchhoff's current law is enforced everywhere on the structure. The input data specify the frequency, wire radius, wire conductivity, the parameters of the exterior medium, coordinates of points to describe the shape and size of the wire configuration, and a list of the wire segments. If some or all of the wire segments are insulated, the radius and permittivity of the insulating sleeve are indicated.

Coordinates are required for wire endpoints, corners, junctions and terminals. For accuracy, the longest wire segment should not greatly exceed one-quarter wavelength. Longer segments should be subdivided by defining additional current-sampling points. The program automatically defines a set of N sinusoidal dipole modes on the wire structure and computes the mutual impedance matrix for these modes. The elements in the matrix are generated by numerical integration when appropriate, or from closed-form expressions in terms of exponential integrals. The computer program uses certain approximations which yield a symmetric matrix even when the wire structure has finite conductivity, lumped loads and insulating sleeves.

In antenna problems, the output data includes the current distribution, impedance, radiation efficiency, gain, patterns and near-zone field. In bistatic scattering problems, the output includes the echo

area and the complex elements of the polarization scattering matrix. In backscatter situations, the output includes also the absorption, scattering and extinction cross sections.

If the wire has finite conductivity or dielectric sleeves, it is assumed that the frequency is real. This restriction can readily be removed if the user will specify the surface impedance of the wire and the complex permittivities of the dielectric sleeves and the ambient medium appropriate for complex frequencies.

The user may make a tradeoff between accuracy and computation costs by specifying the input variable INT. A large value increases the accuracy and the cost. For most problems, the recommended value is $INT = 4$.

The program was run on an IBM 370/165 computer to determine the broadside backscatter for a wire-grid square plate with edge length L . With a five-by-five grid, there are 60 segments, 36 points and 84 simultaneous equations. With $INT = 4$, calculations were made for $L/\lambda = 0.3, 0.4, 0.5, 0.6$ and 0.7 . The execution time was 100 seconds. This averages to 20 seconds for each value of L/λ . The wire structure was perfectly conducting, uninsulated and located in free space. No advantage was taken of the target symmetries.

The next section presents the thin-wire computer program, instructions for the user, typical input and output data and tables of the mutual impedance of sinusoidal dipoles. Appendices list the computer subroutines and explain their functions.

II. THE THIN-WIRE COMPUTER PROGRAM

Fig. 1 is a Fortran listing of the thin-wire computer program. Near the beginning of this program, the DIMENSION statements reserve storage for a wire structure with up to 50 segments, 55 points and 60 dipole modes. Quantities with the same or related dimensions are grouped together on the same line or consecutive lines.

NM denotes the actual number of monopoles (segments), INM is the corresponding dimension, and the dimension for CG, VG and ZLD is twice INM. The second subscript for MD always has a dimension of 4.

N denotes the number of simultaneous linear equations and ICJ is the corresponding dimension. The dimension for C is $(ICJ*ICJ + ICJ)/2$.

The DO LOOP ending at statement 15 sets $ISC(J) = 0$ for all the segments. This indicates that the wires are bare or uninsulated. If some or all of the segments are insulated, the user may set $ISC(J) = 1$ for the appropriate segment numbers J.

```

COMPLEX EP2,EP3,ETA,GAM,Y11,Z11,ZS      0001
COMPLEX EPPS,EPTS,ETPS,ETTS,EX,EY,EZ   0002
COMPLEX C(1830),CJ(60),EP(60),ET(60),EPP(60),ETT(60) 0003
DIMENSION I1(60),I2(60),I3(60),JA(60),JB(60) 0004
COMPLEX CGD(50),SGD(50),CG(100),VG(100),ZLD(100) 0005
DIMENSION D(50),IA(50),IB(50),ISC(50),MD(50,4),ND(50) 0006
DIMENSION X(55),Y(55),Z(55)           0007
DATA PI,TP/3.14159,6.28318/           0008
DATA EO,UO/8.854E-12,1.2566E-6/      0009
2   FORMAT(1X,8F15.7)                  0010
3   FORMAT(1X,4F15.7/)                 0011
4   FORMAT(1X,115,8F14.6)              0012
5   FORMAT(1H0)                         0013
6   FORMAT(1X,6F15.7/)                 0014
7   FORMAT(8F10.5)                      0015
8   FORMAT(1X,114,13I5)                 0016
9   FORMAT(3X,'MAX = ',15,3X,'MIN = ',15,3X,'N = ',15) 0017
    ICJ=60                               0018
    INM=50                               0019
    DO 15 J=1,INM                        0020
15  ISC(J)=0                             0021
    READ(5,7)BM,ER2,SIG2,TD2             0022
    WRITE(6,2)BM,ER2,SIG2,TD2           0023
    READ(5,7)AM,CMM,ER3,SIG3,TD3        0024
    WRITE(6,2)AM,CMM,ER3,SIG3,TD3       0025
    READ(5,8)IBISC,IGAIN,INEAR,ISCAT,IWR,NGEN,NM,NP 0026
    WRITE(6,8)IBISC,IGAIN,INEAR,ISCAT,IWR,NGEN,NM,NP 0027
    READ(5,7)FMC,PHA,THA,PHI,THI,PHS,THS 0028
    WRITE(6,2)FMC,PHA,THA,PHI,THI,PHS,THS 0029
    DO 22 J=1,NM                          0030
    READ(5,8)IA(J),IB(J)                 0031
22  WRITE(6,8)J,IA(J),IB(J)             0032
    DO 40 I=1,NP                          0033
    READ(5,7)X(I),Y(I),Z(I)             0034
40  WRITE(6,4)I,X(I),Y(I),Z(I)         0035
    READ(5,7)XP,YP,ZP                    0036
    FHZ=FMC*1.E6                          0037
    OMEGA=TP*FHZ                          0038
    IF(SIG2.LT.0.)EP2=ER2*EO*CMPLX(1.,-TD2) 0039
    IF(TD2.LT.0.)EP2=CMPLX(ER2*EO,-SIG2/OMEGA) 0040
    IF(SIG3.LT.0.)EP3=ER3*EO*CMPLX(1.,-TD3) 0041
    IF(TD3.LT.0.)EP3=CMPLX(ER3*EO,-SIG3/OMEGA) 0042
    ETA=CSQRT(UO/EP3)                     0043
    GAM=OMEGA*CSQRT(-UO*EP3)              0044
    CALL SORT(IA,IB,I1,I2,I3,JA,JB,MD,ND,NM,NP,N,MAX,MIN,ICJ,INM) 0045
    WRITE(6,5)                             0046
    WRITE(6,9)MAX,MIN,N                    0047
    WRITE(6,5)                             0048
    IF(MAX.GT.4 .OR. MIN.LT.1 .OR. N.GT.ICJ)GO TO 800 0049
    INT=4                                  0050
    I12=1                                  0051
    DO 60 J=1,NM                          0052
    VG(J)=(.0,.0)                          0053
    ZLD(J)=(.0,.0)                         0054
    JJ=J+NM                                0055
    VG(JJ)=(.0,.0)                         0056
60  ZLD(JJ)=(.0,.0)                       0057
    IF(NGEN.GT.0)VG(NGEN)=(1.,0.)         0058
    CALL SGANT(IA,IB,INM,INT,ISC,I1,I2,I3,JA,JB,MD,N,ND,NM,NP 0059
2,AM,BM,C,CGD,CMM,D,EP2,EP3,ETA,FHZ,GAM,SGD,X,Y,Z,ZLD,ZS) 0060
    IF(N.LE.0)GO TO 800                    0061
    IF(NGEN.LE.0)GO TO 400                0062

```

Fig. 1a. The thin-wire computer program.

CALL GANT1(IA,IB,INM,IWR,I1,I2,I3,I12,JA,JB,MD,N,ND,NM,AM	0063
2,C,CJ,CG,CMM,D,EFF,GAM,GG,CGD,SGD,VG,Y11,Z11,ZLD,ZS)	0064
WRITE(6,3)EFF,GG,Z11	0065
200 IF(INEAR.LE.0)GO TO 300	0066
CALL GNFLD(IA,IB,INM,I1,I2,I3,MD,N,ND,NM,AM,CGD,SGD,ETA,GAM	0067
2,CJ,D,X,Y,Z,XP,YP,ZP,EX,EY,EZ)	0068
WRITE(6,3)XP,YP,ZP	0069
WRITE(6,6)EX,EY,EZ	0070
300 IF(IGAIN.LE.0)GO TO 400	0071
INC=0	0072
PH=PHA	0073
TH=THA	0074
CALL GFFLD(IA,IB,INC,INM,IWR,I1,I2,I3,I12,MD,N,ND,NM,AM	0075
2,ACSP,ACST,C,CGD,CG,CJ,CMM,D,ECSP,ECST,EP,ET,EPP,ETT,EPPS,EPTS	0076
3,ETPS,ETTS,GG,GPP,GTT,PH,SGD,SCSP,SCST,SPPM,SPTM,STPM,STTM,TH	0077
4,X,Y,Z,ZLD,ZS,ETA,GAM)	0078
WRITE(6,3)PH,TH,GPP,GTT	0079
400 IF(ISCAT.LE.0)GO TO 600	0080
INC=1	0081
PH=PHI	0082
TH=THI	0083
CALL GFFLD(IA,IB,INC,INM,IWR,I1,I2,I3,I12,MD,N,ND,NM,AM	0084
2,ACSP,ACST,C,CGD,CG,CJ,CMM,D,ECSP,ECST,EP,ET,EPP,ETT,EPPS,EPTS	0085
3,ETPS,ETTS,GG,GPP,GTT,PH,SGD,SCSP,SCST,SPPM,SPTM,STPM,STTM,TH	0086
4,X,Y,Z,ZLD,ZS,ETA,GAM)	0087
WRITE(6,6)PH,TH,SPPM,SPTM,STPM,STTM	0088
WRITE(6,6)ACSP,ACST,ECSP,ECST,SCSP,SCST	0089
500 IF(IBISC.LE.0)GO TO 600	0090
INC=2	0091
PH=PHS	0092
TH=THS	0093
CALL GFFLD(IA,IB,INC,INM,IWR,I1,I2,I3,I12,MD,N,ND,NM,AM	0094
2,ACSP,ACST,C,CGD,CG,CJ,CMM,D,ECSP,ECST,EP,ET,EPP,ETT,EPPS,EPTS	0095
3,ETPS,ETTS,GG,GPP,GTT,PH,SGD,SCSP,SCST,SPPM,SPTM,STPM,STTM,TH	0096
4,X,Y,Z,ZLD,ZS,ETA,GAM)	0097
WRITE(6,6)PH,TH,SPPM,SPTM,STPM,STTM	0098
600 CONTINUE	0099
800 CALL EXIT	0100
END	

Fig. 1b. The thin-wire computer program.

The first READ statement inputs the following parameters for the dielectric insulation:

BM	outer radius in meters
ER2	dielectric constant relative to free space
SIG2	conductivity in mhos per meter
TD2	loss tangent

The program will use SIG2 or TD2 but not both. The user determines which one will be used by assigning the other a negative value. For an uninsulated wire structure, the program will not use any of the data from the first READ statement.

The second READ statement inputs the following parameters for the wire and the exterior medium:

AM	wire radius in meters
CMM	wire conductivity in megamhos per meter
ER3	dielectric constant relative to free space
SIG3	conductivity in mhos per meter
TD3	loss tangent

The parameters ER3, SIG3 and TD3 are those of the homogeneous ambient medium. Again, the program will use SIG3 or TD3 but not both.

The third READ statement inputs the following data:

IBISC	indicator for bistatic scattering calculations
IGAIN	indicator for antenna gain calculations
INEAR	indicator for near-zone field calculations
ISCAT	indicator for backscatter calculations
IWR	indicator for writeout of current distributions
NGEN	indicator for antenna calculations
NM	number of monopoles (segments)
NP	number of points

For each indicator, a positive value means the calculation or writeout is desired while a zero or negative value means it is not desired.

The fourth READ statement inputs the following data:

FMC	frequency in megahertz
PHA,THA	far-field angle for antenna gain
PHI,THI	incidence angle for plane-wave scattering
PHS,THS	scattering angle for bistatic scattering

The above angles are given in degrees, and they denote values of the angular coordinates in the spherical system (r, θ, ϕ) widely used in antenna and scattering literature.

The fifth READ statement (in the DO LOOP ending with statement 22) inputs the endpoints IA(J) and IB(J) of segment J. Thus, IA and IB are the index numbers of the two points which are joined by segment J.

The sixth READ statement (in the DO LOOP ending with statement 40) inputs the coordinates X(I), Y(I) and Z(I) of point I in meters. The seventh and last READ statement inputs the coordinates XP, YP and ZP (in meters) of the observation point for near-zone field calculations.

Some of the quantities used in the program are defined as follows:

FHZ	frequency in Hertz
OMEGA	angular frequency
EP2	complex permittivity of insulation
EP3	complex permittivity of ambient medium
ETA	intrinsic impedance of ambient medium
GAM	intrinsic propagation constant of ambient medium
ZS	surface impedance of wire

For an uninsulated wire with perfect conductivity, one may specify complex values for ETA and GAM and delete the following input data and calculations: BM, ER2, SIG2, TD2, ER3, SIG3, TD3, FMC, FHZ, OMEGA, EP2 and EP3.

After reading the input data, the program calls subroutine SORT. This subroutine defines a set of dipole modes on the wire structure. N denotes the total number of dipole modes, the number of simultaneous linear equations, and the size of the impedance matrix Z_{ij} . Since this matrix is symmetric, only the upper-right triangular portion (including the entire principal diagonal) is calculated and stored in C(K). SORT calculates N, but the user may predict N as follows to reserve adequate storage. If m wire segments intersect at a point, this point is defined as a junction of order m and degree $n = m - 1$. There will be n dipole modes with terminals at this junction. N is determined by summing the degrees of all the junctions. For an example, an endpoint of a dipole is a junction of order $m = 1$ and degree $n = 0$. The vertex of a V dipole is a junction of order 2 and degree 1. NP denotes the number of points on the wire structure, and each of these points is considered to be a junction.

Mode I is a two-segment V dipole with a sinusoidal current distributed over the intersecting segments JA(I) and JB(I). The dipole has endpoints I1(I) and I3(I) and terminals at I2(I). The reference direction for positive current on dipole I is from I1 to I2 to I3.

A wire segment may be shared by as many as four dipole modes, or as few as one. In the output of subroutine SORT, ND(J) denotes the number of dipoles sharing segment J. The extreme values of ND(J) are MAX and MIN. If MIN is less than one, the wire structure has an unconnected segment and the computation is aborted. (An isolated wire

must have at least two segments and three points.) If N exceeds ICJ, the dimensions are inadequate and the run is aborted.

INT specifies the number of intervals for calculating the elements in the impedance matrix with Simpson's-rule integration. A large value for INT improves the accuracy at the expense of greater execution time. For most problems a suitable combination of speed and accuracy is obtained with INT = 4. A larger value is recommended if one wire passes close to another as in the helix or the multiturn loop. If in doubt, one may set INT = 0 to choose the rigorous closed-form impedance expressions in terms of exponential integrals.

The DO LOOP ending with statement 60 sets all the lumped load impedances and generator voltages to zero. If the wire structure has lumped loads, one may insert a READ command after statement 60 to input a list of complex load impedances ZLD(J). For a wire antenna with just one generator, the program inserts a unit voltage generator with VG(NGEN) = (1.,0.). If the antenna or array has several generators, one may insert a READ command after statement 60 to input a list of complex voltages VG(J).

Generators or lumped loads may be inserted at either end or both ends of segment J. First consider a load impedance inserted in the middle of segment J. Now slide the load along the segment and let it approach the endpoint IA(J). This load is represented by ZLD(J). Next insert another load in segment J and slide it to approach the endpoint IB(J). This load is designated ZLD(JJ) where JJ = J + NM. The same convention is employed for the voltage generators VG(J) and VG(JJ). A generator voltage VG(J) is considered positive if it tends to force a current flow in the direction from IA(J) to IB(J).

Subroutine SGANT calculates the elements of the impedance matrix Z_{ij} and stores them in C(K) where $K = (I-1)*N - (I*I - I)/2 + J$. This subroutine will set N = 0 and the run will abort if the wire radius is zero or negative, the shortest segment length is less than the wire diameter, the wire radius is electrically large, or the longest segment is too long.

Subroutine GANT1 considers the thin-wire structure as an antenna and solves for the current distribution CG(J), radiation efficiency EFF, time-average power input GG and complex power input Y11. In the current distribution, CG(J) is the current on segment J as one approaches the endpoint IA(J) and CG(JJ) is the current at the other end IB(J). The reference direction for positive current is from IA to IB. Thus, the conventions are the same for the branch currents CG and the branch voltages VG.

If the antenna has only one voltage generator with VG(NGEN) = (1.,0.), then Y11 is the antenna admittance and Z11 is the impedance.

The radiation efficiency EFF is calculated from the time-average power input to the antenna and the time-average power dissipated in the wire and the lumped loads. If the antenna is insulated, the power dissipated in the insulation is neglected. If the wire has perfect conductivity and the loads are purely reactive, the calculated efficiency will be 100 per cent.

The near-field subroutine GNFLD calculates the electric field intensity (EX,EY,EZ) at the observation point (XP,YP,ZP). In the calling parameters, CJ denotes the current distribution on the wire. (The loop currents are stored in CJ(I) and the branch currents in CG(J)). Thus, the currents must be calculated before GNFLD is called. Fig. 1 illustrates the use of GNFLD to calculate the near-zone field in an antenna problem. This subroutine can be called again just above statement 500 to calculate the near-zone scattered field for a wire target. In the calling parameters, CJ is replaced with EP or ET to obtain the near-zone field with a phi-polarized or theta-polarized incident plane wave. Reference 1 describes the more sophisticated techniques required when the observation point is extremely close to the wire structure.

The far-field subroutine GFFLD calculates antenna gain if INC = 0, backscattering if INC = 1, and bistatic scattering if INC = 2. If INC = 0, PH and TH denote the spherical coordinates ϕ and θ of the distant observation point and the output from GFFLD is defined as follows. EPPS and ETTS denote the phi-polarized and theta-polarized components of the electric field intensity. For example,

$$(1) \quad EPPS = r e^{\gamma r} E_{\phi}$$

where r is the distance from the origin to the observation point. GPP and GTT denote the power gains associated with the phi and theta polarizations. Appendix 14 defines GPP and GTT more precisely.

If INC = 1, PH and TH denote the incidence angles ϕ_i and θ_i . These are also the spherical coordinates of the distant source. In this backscattering situation, the output data from GFFLD are defined as follows:

ACSP,ACST	absorption cross sections for ϕ and θ polarizations
ECSP,ECST	extinction cross sections for ϕ and θ polarizations
EP,ET	loop currents induced by ϕ and θ polarized waves
EPPS	scattered electric field $E_{\phi\phi}$
EPTS	scattered electric field $E_{\phi\theta}$
ETPS	scattered electric field $E_{\theta\phi}$
ETTS	scattered electric field $E_{\theta\theta}$
SCSP,SCST	scattering cross sections for ϕ and θ polarizations
SPPM	echo area $\sigma_{\phi\phi}$

SPTM	echo area $\sigma_{\phi\theta}$
STPM	echo area $\sigma_{\theta\phi}$
STTM	echo area $\sigma_{\theta\theta}$

The echo areas are given in square meters. For the doubly-subscripted quantities such as $E_{\phi\phi}$ and $\sigma_{\phi\phi}$, the first and second subscripts specify the polarizations of the incident and scattered waves, respectively. The complex numbers EPPS, EPTS, ETPS and ETTS are the elements of the polarization scattering matrix.

If INC = 2, PH and TH denote the scattering angles ϕ_s and θ_s . These are the spherical coordinates of the distant observer. In this bistatic scattering situation, the only outputs from GFFLD are the polarization scattering matrix and the echo areas.

To obtain antenna patterns, backscattering patterns or bistatic patterns, one may insert DO LOOPS in the program to increment the angles PH and TH. The DO LOOP will begin just above the call to GFFLD and terminate just below this call. To obtain the near-zone field distribution along a given probing path, one may insert a DO LOOP beginning just above the call to GNFLD and terminating just below this call.

When the calculations have been completed for one problem, one may GO TO a point just above CALL GANT1 if only the generator voltages are to be changed. One may GO TO a point just below CALL SORT if there is a change in the wire radius or conductivity, the insulation, ambient medium, frequency, load impedances or the coordinates (X,Y,Z). If there is a change in NM, NP, IA or IB, one should GO TO a point above CALL SORT.

Consider an array of three center-fed dipoles, and suppose we desire the 3 x 3 admittance matrix for the array. Let each dipole be divided into four segments with segments 1 through 4 on dipole 1, 5 through 8 on dipole 2 and 9 through 12 on dipole 3. The three-port admittance matrix can be obtained by inserting a DO LOOP beginning just above CALL GANT1 and terminating just below this call. GANT1 will be called three times with all the voltages VG set to zero except for a single one-volt generator. On the first, second and third calls, let NGEN = 3, 7 and 11 to represent a generator at the center of dipole 1, 2 and 3, respectively. After the first call, set $Y_{11} = CG(3)$, $Y_{12} = CG(7)$ and $Y_{13} = CG(11)$. Set $Y_{22} = CG(7)$ and $Y_{23} = CG(11)$ after the second call and $Y_{33} = CG(11)$ after the third call.

For extremely small antennas, quasi-static or double-precision subroutines are required.

The wire radius must exceed zero, but there is no difficulty with small radii. If the radius exceeds 0.007λ , the thin-wire assumptions are questionable and the accuracy and convergence deteriorate. The length ratio of the longest and shortest segments should not exceed 100. It is

assumed that the wire length exceeds the wire diameter by a factor of at least 30. We are not aware of any lower limit on the segment length, however.

If a wire is bent sharply to form a small acute angle (less than 30 degrees), the thin-wire model is questionable. It is assumed that the wire conductivity greatly exceeds the conductivity of the ambient medium. For insulated wires, the dielectric layer is assumed to be electrically thin.

For each thin-wire problem, calculations should be repeated several times with the wire divided progressively into shorter segments. There is no assurance of accuracy until the output data converge. For a moderately thick wire (with radius $a = 0.007 \lambda$ or larger), the susceptance may diverge with the delta-gap model. This difficulty may be alleviated or eliminated with the magnetic-frill model and the techniques of Imbriale and Ingerson [2].

Tables 1, 2 and 3 list input and output data for three simple examples of uninsulated wire structures. Each table includes a sketch of the wire configuration with labels to indicate the numbering system for the points and segments. In these examples there are no lumped loads.

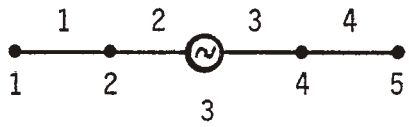
In the sinusoidal-reaction formulation, a basic function is the mutual impedance between two sinusoidal elementary electric dipoles. One dipole is a test source located on the axis of the wire structure, and the other is an expansion mode on the wire surface. In view of the importance of this mutual impedance, short tables are presented next for a few simple cases. The data can be reproduced with the program in Fig. 1 with appropriate input data for uninsulated wires with perfect conductivity and no lumped loads in free space. The data were obtained with the closed-form expressions ($INT = 0$) by writing out the quantities $C(K)$ just below the call to subroutine SGANT. Double precision was used for these calculations.

Table 4 lists the self impedance of a two-segment sinusoidal V dipole with radius $a = 0.001 \lambda$. Subroutine SGANT calculates this quantity by setting up one elementary dipole on the wire axis and another identical dipole on the wire surface. These dipoles lie in parallel planes separated by a distance equal to the wire radius.

In Table 5, dipoles 1 and 2 have terminals at vertices 1 and 2, respectively, and they share the middle segment. Again these dipoles lie in parallel planes separated by a distance equal to the wire radius. For a one-turn planar polygon wire loop, subroutine SGANT would generate the data in Table 4 for the diagonal elements Z_{ij} and the data in Table 5 for the next elements.

TABLE 1
Input and Output Data for Straight Wire

Input Data								
0.002	2.56	-1.0	0.0005					
0.001	1.00	1.0	-1.0	0.0				
1	1	1	1	0	3	4	5	
300.	0.	90.	0.	90.	45.	45.		
1	2							
2	3							
3	4							
4	5							
0.	0.	-0.250						
0.	0.	-0.125						
0.	0.	0.						
0.	0.	0.125						
0.	0.	0.250						
1.	1.	1.						

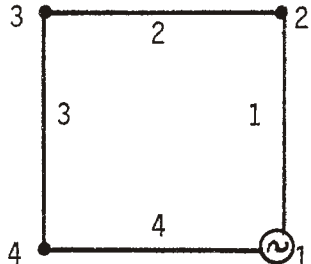


The diagram shows a horizontal line representing a straight wire with five nodes labeled 1, 2, 3, 4, and 5 from left to right. A current source, represented by a circle with a tilde symbol (~), is connected between nodes 2 and 3.

Output Data					
98.18	0.0095	82.97	43.26		
-.091	0.080	-0.091	0.080	0.224	-0.096
0.0	90.0	0.0	1.615		
0.0	90.0	0.0	0.0	0.0	0.608
0.0	0.0069	0.0	0.377	0.0	0.370
45.0	45.0	0.0	0.0	0.0	0.239

TABLE 2
Input and Output Data for Square Loop

Input Data							
0.002	2.56	-1.0	0.0005				
0.001	1.0	1.0	-1.0	0.0			
1	1	1	1	0	1	4	4
300.	0.0	90.0	0.0	90.0	45.	45.	
1	2						
2	3						
3	4						
4	1						
0.05	-0.05	0.0					
0.05	0.05	0.0					
-0.05	0.05	0.0					
-0.05	-0.05	0.0					
1.0	1.0	1.0					



The diagram shows a square loop with four nodes labeled 1, 2, 3, and 4. Node 1 is at the bottom right, node 2 is at the top right, node 3 is at the top left, and node 4 is at the bottom left. A current source, represented by a circle with a tilde symbol (~), is connected between nodes 1 and 4.

Output Data					
73.10	.243E-4	62.94	1609.8		
-.0078	.0027	.0057	.0029	-.0010	-.0056
0.0	90.0	.8066	.0		
0.0	90.0	.0002	.0	.0	.0
.126E-4	0.0	.936E-4	.0	.810E-4	.0
45.0	45.0	.106E-3	.265E-4	.0	.0

TABLE 3
Input and Output Data for Y Antenna

Input Data						
0.002	2.56	-1.0	0.0005			
0.001	1.0	1.0	-1.0	0.0		
1	1	1	1	0	2	4
300.	0.0	90.0	0.0	90.0	45.	45.
1	2					5
2	3					
3	4					
3	5					
0.0	-0.30	0.0				
0.0	-0.15	0.0				
0.0	0.0	0.0				
0.1	0.1	0.0				
-0.1	0.1	0.0				
1.0	1.0	1.0				

Output Data						
97.88	0.013	75.53	-0.572			
-.124	0.081	0.260	-0.064	-0.126	0.070	
0.0	90.0	1.535	0.0			
0.0	90.0	0.748	0.0	0.0	0.0	
0.0103	0.0	0.487	0.0	0.477	0.0	
45.0	45.0	0.360	0.170	0.0	0.0	

TABLE 4

Self Impedance of Two-Segment V Dipole Shown in Fig. 2
Radius: $a = 0.001\lambda$

ψ	$h = 0.10\lambda$	$h = 0.15\lambda$	$h = 0.20\lambda$	$h = 0.25\lambda$
30°	0.59 - j 481	1.4 - j 314	3.1 - j 186	6.1 - j 61
60	2.15 - j 547	5.3 - j 337	11.0 - j 177	21.3 - j 21
90	4.22 - j 572	10.4 - j 340	21.1 - j 163	40.0 + j 9
120	6.31 - j 583	15.3 - j 338	30.9 - j 151	57.7 + j 28
150	7.81 - j 587	18.9 - j 335	37.7 - j 144	69.3 + j 39
180	8.33 - j 589	20.1 - j 335	39.9 - j 142	73.1 + j 42

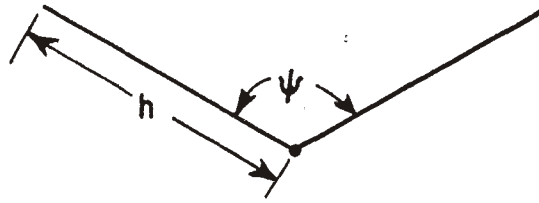


Fig. 2. Symmetric center-fed V dipole.

TABLE 5

Mutual Impedance Between Overlapping V Dipoles in Fig. 3
 Radius: $a = 0.001\lambda$

ψ	$h = 0.10\lambda$	$h = 0.15\lambda$	$h = 0.20\lambda$	$h = 0.25\lambda$
60°	$-0.96 + j 338$	$-2.08 + j 285$	$-3.45 + j 275$	$-4.8 + j 298$
90	$0.19 + j 322$	$1.03 + j 276$	$3.57 + j 271$	$10.1 + j 297$
120	$3.29 + j 336$	$8.40 + j 290$	$17.86 + j 285$	$35.3 + j 309$
150	$6.61 + j 346$	$15.61 + j 299$	$30.00 + j 291$	$52.9 + j 309$
180	$8.01 + j 349$	$18.47 + j 301$	$34.35 + j 292$	$58.2 + j 308$

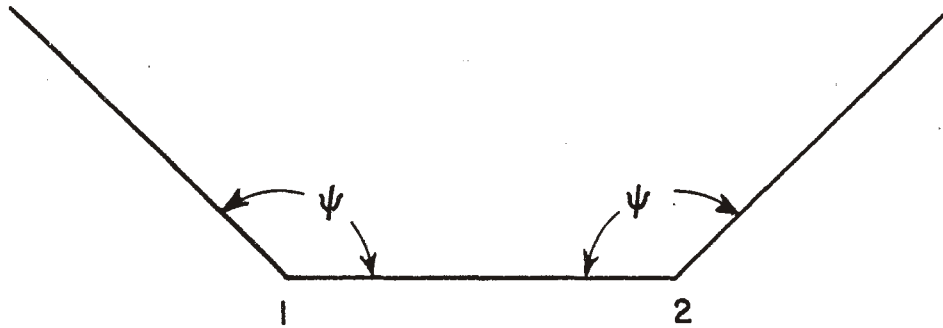


Fig. 3. Overlapping V dipoles share the middle segment.

Tables 6, 7, and 8 list the mutual impedance for other configurations. In all these tables, the data apply to two-segment center-fed sinusoidal dipoles with identical segment lengths h .

TABLE 6
 Mutual Impedance Between Overlapping V Dipoles in Fig. 4
 Radius: $a = 0.001\lambda$

α	$h = 0.10\lambda$	$h = 0.15\lambda$	$h = 0.20\lambda$	$h = 0.25\lambda$
30°	6.74 - j 314	16.24 - j 167	32.17 - j 56	58.7 + j 49.6
60	3.16 - j 291	7.68 - j 169	15.47 - j 76	28.8 + j 14.2
90	0.06 - j 278	0.31 - j 172	1.15 - j 92	3.5 - j 12.2
120	-1.01 - j 256	-2.39 - j 168	-4.47 - j 101	-7.6 - j 35.5
150	-0.48 - j 207	-1.20 - j 146	-2.40 - j 98	-4.5 - j 50.7

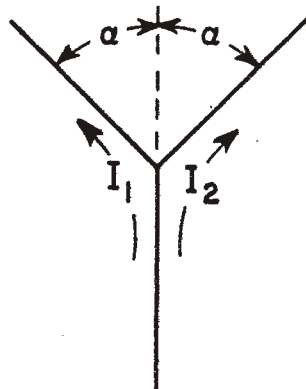


Fig. 4. Overlapping V dipoles share the bottom segment in a planar Y configuration.

TABLE 7
 Mutual Impedance Between the Coplanar-Skew Linear Dipoles in Fig. 5
 Displacement: $d = \lambda$

θ	$h = 0.10\lambda$	$h = 0.15\lambda$	$h = 0.20\lambda$	$h = 0.25\lambda$
0°	$0.337 + j 1.952$	$0.880 + j 4.759$	$1.932 + j 9.547$	$4.011 + j 17.7$
15	$0.322 + j 1.884$	$0.831 + j 4.585$	$1.799 + j 9.180$	$3.671 + j 17.0$
30	$0.281 + j 1.684$	$0.700 + j 4.082$	$1.448 + j 8.128$	$2.800 + j 15.0$
45	$0.220 + j 1.369$	$0.521 + j 3.301$	$1.000 + j 6.519$	$1.745 + j 11.9$
60	$0.149 + j 0.964$	$0.333 + j 2.310$	$0.579 + j 4.524$	$0.860 + j 8.1$
75	$0.075 + j 0.497$	$0.159 + j 1.187$	$0.252 + j 2.308$	$0.305 + j 4.1$
90	$0.0 + j 0.0$	$0.0 + j 0.0$	$0.0 + j 0.0$	$0.0 + j 0.0$

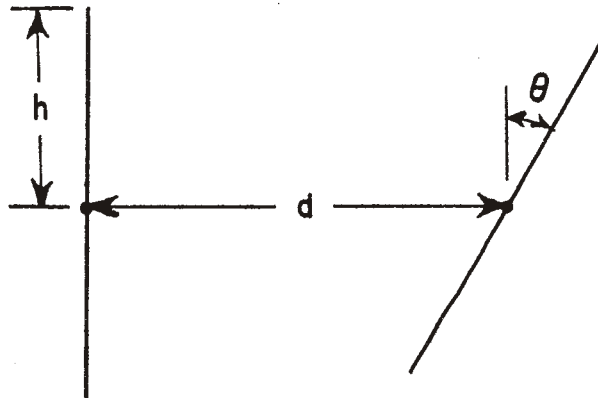


Fig. 5. Center-fed coplanar-skew linear dipoles.

TABLE 8
 Mutual Impedance Between the Nonplanar-Skew Linear Dipoles in Fig. 6
 Displacement: $d = \lambda$

ϕ	$h = 0.10\lambda$	$h = 0.15\lambda$	$h = 0.20\lambda$	$h = 0.25\lambda$
0°	0.337 + j 1.952	0.880 + j 4.759	1.932 + j 9.547	4.011 + j 17.74
15	0.326 + j 1.886	0.850 + j 4.596	1.867 + j 9.222	3.877 + j 17.14
30	0.292 + j 1.691	0.762 + j 4.121	1.675 + j 8.269	3.482 + j 15.37
45	0.238 + j 1.380	0.622 + j 3.365	1.369 + j 6.752	2.850 + j 12.55
60	0.169 + j 0.976	0.440 + j 2.380	0.969 + j 4.775	2.020 + j 8.88
75	0.087 + j 0.505	0.228 + j 1.232	0.502 + j 2.472	1.047 + j 4.60
90	0.0 + j 0.0	0.0 + j 0.0	0.0 + j 0.0	0.0 + j 0.0

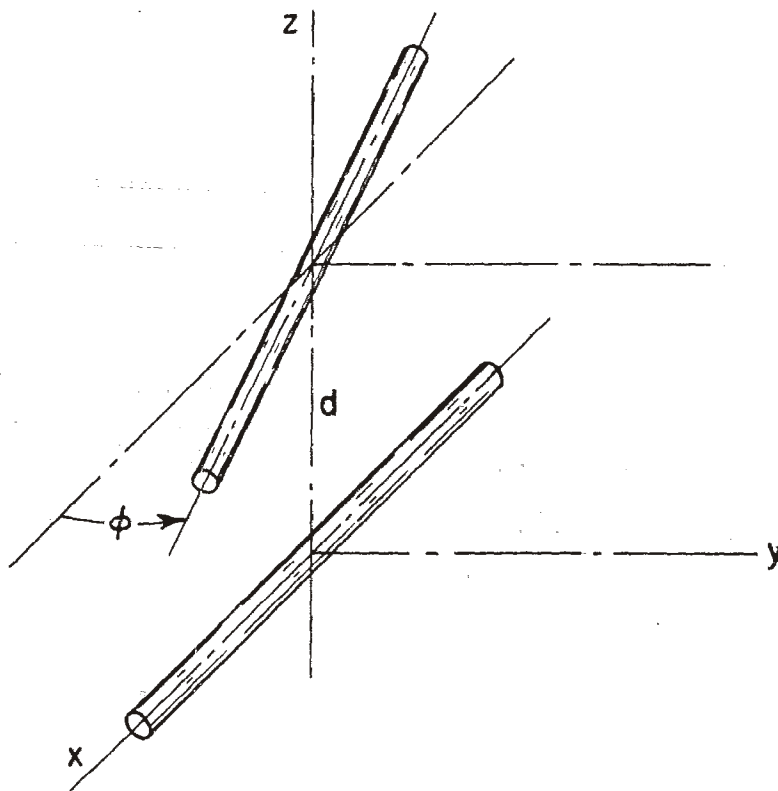


Fig. 6. Center-fed nonplanar-skew linear dipoles.

III. SUMMARY

This report presents the sinusoidal-reaction computer program for thin-wire antennas and scatterers, instructions for the user, typical input and output data and mutual-impedance tables for sinusoidal dipoles. Appendices list the computer subroutines and explain their functions.

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1. Richmond, J. H., "Radiation and Scattering by Thin-Wire Structures in the Complex Frequency Domain", Interaction Note 202, May 1974.
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3. Abramowitz, M., and Stegun, I.A., "Handbook of mathematical functions with formulas, graphs, and mathematical tables," National Bureau of Standards, Applied Mathematics Series AMS-55, 1964, Chapter 5.
4. Faddeev, D.K., and Faddeeva, V.N., Computational Methods of Linear Algebra, W. H. Freeman and Company, San Francisco, 1963, pp. 144-147.

APPENDIX 1. Subroutine SORT

Subroutine SORT, listed in Fig. 7, defines a set of dipole mode currents on the wire structure. The input data IA, IB, NM, NP, ICJ and INM have been defined already. The output data are defined as follows:

N	total number of dipole modes
I1(I)	endpoint of dipole I
I2(I)	terminal point of dipole I
I3(I)	endpoint of dipole I
JA(I)	first segment of dipole I
JB(I)	second segment of dipole I
MD(J,K)	list of dipoles sharing segment J
ND(J)	total number of dipoles sharing segment J
MAX,MIN	extreme values of ND(J)

At completion of the DO LOOP ending with statement 20, NJK denotes the number of segments intersecting at point K, and JSP is a list of these segments. In the DO LOOP ending with statement 22, the computer sets up the appropriate number MOD of dipole modes with terminals at point K.

APPENDIX 2. Subroutine SGANT

Subroutine SGANT, listed in Fig. 8, calculates the mutual impedances Z_{ij} and stores them in C(K). The input data for SGANT have been defined already. The output data are defined as follows:

C(K)	open-circuit impedance matrix
CGD(J)	cosh γd for segment J
SGD(J)	sinh γd for segment J
D(J)	length of segment J
ZS	surface impedance of the wire

The surface impedance is calculated just above statement 12. B01 denotes J_0/J_1 where J_0 and J_1 are the Bessel functions of order zero and one with complex argument ZARG. It is assumed that all the wire segments have the same radius, conductivity and surface impedance.

In the DO LOOP ending with statement 20, SGANT calculates the segment lengths D(J). DMIN and DMAX denote the lengths of the shortest and longest segments. If the wire radius or the segment lengths are clearly beyond the range of thin-wire theory, N is set to zero at statement 25 followed by RETURN to the main program to abort the calculation.

At statement 30, the program selects a segment K, and a few statements below this it selects another segment L. K is a segment of test dipole I, and L is a segment of expansion mode J. The mutual impedance between segments K and L is obtained by calling subroutine GGS or GGMM.

	SUBROUTINE SORT(IA,IB,I1,I2,I3,JA,JB,MD,ND,NM,NP,N,MAX,MIN	0001
	2,ICJ,INM)	0002
	DIMENSION JSP(20)	0003
	DIMENSION I1(1),I2(1),I3(1),JA(1),JB(1)	0004
	DIMENSION IA(1),IB(1),ND(1),MD(INM,4)	0005
	I=0	0006
	DO 24 K=1,NP	0007
	NJK=0	0008
	DO 20 J=1,NM	0009
	IND=(IA(J)-K)*(IB(J)-K)	0010
	IF(IND.NE.0)GO TO 20	0011
	NJK=NJK+1	0012
	JSP(NJK)=J	0013
20	CONTINUE	0014
	MOD=NJK-1	0015
	IF(MOD.LE.0)GO TO 24	0016
	DO 22 IMD=1,MOD	0017
	I=I+1	0018
	IF(I.GT.ICJ)GO TO 22	0019
	IPD=IMD+1	0020
	JAI=JSP(IMD)	0021
	JA(I)=JAI	0022
	JBI=JSP(IPD)	0023
	JB(I)=JBI	0024
	I1(I)=IA(JAI)	0025
	IF(IA(JAI).EQ.K)I1(I)=IB(JAI)	0026
	I2(I)=K	0027
	I3(I)=IA(JBI)	0028
	IF(IA(JBI).EQ.K)I3(I)=IB(JBI)	0029
22	CONTINUE	0030
24	CONTINUE	0031
	N=I	0032
	DO 30 J=1,NM	0033
	ND(J)=0	0034
	DO 30 K=1,4	0035
30	MD(J,K)=0	0036
	III=N	0037
	IF(N.GT.ICJ)III=ICJ	0038
	DO 40 I=1,III	0039
	J=JA(I)	0040
	DO 38 L=1,2	0041
	ND(J)=ND(J)+1	0042
	K=1	0043
	M=0	0044
32	MJK=MD(J,K)	0045
	IF(MJK.NE.0)GO TO 34	0046
	M=1	0047
	MD(J,K)=I	0048
34	K=K+1	0049
	IF(K.GT.4)GO TO 38	0050
	IF(M.EQ.0)GO TO 32	0051
38	J=JB(I)	0052
40	CONTINUE	0053
	MIN=100	0054
	MAX=0	0055
	DO 46 J=1,NM	0056
	NDJ=ND(J)	0057
	IF(NDJ.GT.MAX)MAX=NDJ	0058
46	IF(NDJ.LT.MIN)MIN=NDJ	0059
	RETURN	0060
	END	0061

Fig. 7. Subroutine SORT

```

SUBROUTINE SGANT(IA,IB,INM,INT,ISC,I1,I2,I3,JA,JB,MD,N,ND,NM,NP      0001
2,AM,BM,C,CGD,CMM,D,EP2,EP3,ETA,FHZ,GAM,SGD,X,Y,Z,ZLD,ZS)          0002
COMPLEX ZG,ZH,ZS,EGD,GD,CGDS,SGDS,SGDT,B01                        0003
COMPLEX P11,P12,P21,P22,Q11,Q12,Q21,Q22,EP2,EP,ETA,GAM,EP3       0004
COMPLEX EPSILA,CWEA,BETA,ZARG                                     0005
COMPLEX P(2,2),Q(2,2),CGD(1),SGD(1),C(1),ZLD(1)                 0006
DIMENSION X(1),Y(1),Z(1),D(1),IA(1),IB(1),MD(INM,4)            0007
DIMENSION I1(1),I2(1),I3(1),JA(1),JB(1),ND(1),ISC(1)          0008
DATA EO,TP,UO/8.854E-12,6.28318,1.2566E-6/                      0009
2 FORMAT(3X,'AM = ',E10.3,3X,'DMAX = ',E10.3,3X,'DMIN = ',E10.3) 0010
EP=EP3                                                            0011
ICC=(N*N+N)/2                                                    0012
DO 10 I=1,ICC                                                     0013
10 C(I)=(.0,.0)                                                  0014
ZS=(.0,.0)                                                       0015
IF(CMM.LE.0.)GO TO 12                                           0016
OMEGA=TP*FHZ                                                      0017
EPSILA=CMPLX(EO,-CMM*1.E6/OMEGA)                                0018
CWEA=(.0,1.)*OMEGA*EPSILA                                       0019
BETA=OMEGA*SQRT(UO)*CSQRT(EPSILA-EP)                            0020
ZARG=BETA*AM                                                      0021
CALL CBES(ZARG,B01)                                             0022
ZS=BETA*B01/CWEA                                                0023
12 ZH=ZS/(TP*AM*GAM)                                           0024
DMIN=1.E30                                                       0025
DMAX=.0                                                           0026
DO 20 J=1,NM                                                      0027
K=IA(J)                                                           0028
L=IB(J)                                                           0029
D(J)=SQRT((X(K)-X(L))**2+(Y(K)-Y(L))**2+(Z(K)-Z(L))**2)       0030
IF(D(J).LT.DMIN)DMIN=D(J)                                       0031
IF(D(J).GT.DMAX)DMAX=D(J)                                       0032
EGD=CEXP(GAM*D(J))                                              0033
CGD(J)=(EGD+1./EGD)/2.                                          0034
20 SGD(J)=(EGD-1./EGD)/2.                                        0035
IF(DMIN.LT.2.*AM)GO TO 25                                       0036
IF(CABS(GAM*AM).GT.0.06)GO TO 25                                0037
IF(CABS(GAM*DMAX).GT.3.)GO TO 25                                0038
IF(AM.GT.0.)GO TO 30                                           0039
25 N=0                                                            0040
WRITE(6,2)AM,DMAX,DMIN                                         0041
RETURN                                                            0042
30 DO 200 K=1,NM                                                 0043
NDK=ND(K)                                                         0044
KA=IA(K)                                                         0045
KB=IB(K)                                                         0046
DK=D(K)                                                           0047
CGDS=CGD(K)                                                       0048
SGDS=SGD(K)                                                       0049
DO 200 L=1,NM                                                     0050
NDL=ND(L)                                                         0051
LA=IA(L)                                                         0052
LB=IB(L)                                                         0053
DL=D(L)                                                           0054
SGDT=SGD(L)                                                       0055
NIL=0.                                                            0056
DO 200 II=1,NDK                                                  0057
I=MD(K,II)                                                       0058
MM=(I-1)*N-(I*I-1)/2                                           0059
FI=1.                                                            0060
IF(KB.EQ.I2(II))GO TO 36                                       0061
IF(KB.EQ.I1(II))FI=-1.                                         0062

```

Fig. 8a. Subroutine SGANT

	IS=1	0063
	GO TO 40	0064
36	IF (KA.EQ.I3(I))FI=-1.	0065
	IS=2	0066
40	DO 200 JJ=1,NDL	0067
	J=MD(L,JJ)	0068
	MMM=MM+J	0069
	IF (I.GT.J)GO TO 200	0070
	FJ=1.	0071
	IF (LB.EQ.I2(J))GO TO 46	0072
	IF (LB.EQ.I1(J))FJ=-1.	0073
	JS=1	0074
	GO TO 50	0075
46	IF (LA.EQ.I3(J))FJ=-1.	0076
	JS=2	0077
50	IF (NIL.NE.O)GO TO 168	0078
	NIL=1	0079
	IF (K.EQ.L)GO TO 120	0080
	IND=(LA-KA)*(LB-KA)*(LA-KB)*(LB-KB)	0081
	IF (IND.EQ.O)GO TO 80	0082
C	SEGMENTS K AND L SHARE NO POINTS	0083
	CALL GGS(X(KA),Y(KA),Z(KA),X(KB),Y(KB),Z(KB),X(LA),Y(LA),Z(LA)	0084
	2,X(LB),Y(LB),Z(LB),AM,DK,CGDS,SGDS,DL,SGDT,INT,ETA,GAM	0085
	3,P(1,1),P(1,2),P(2,1),P(2,2))	0086
	GO TO 168	0087
C	SEGMENTS K AND L SHARE ONE POINT (THEY INTERSECT)	0088
80	KG=0	0089
	JM=KB	0090
	JC=KA	0091
	KF=1	0092
	IND=(KB-LA)*(KB-LB)	0093
	IF (IND.NE.O)GO TO 82	0094
	JC=KB	0095
	KF=-1	0096
	JM=KA	0097
	KG=3	0098
82	LG=3	0099
	JP=LA	0100
	LF=-1	0101
	IF (LB.EQ.JC)GO TO 83	0102
	JP=LB	0103
	LF=1	0104
	LG=0	0105
83	SGN=KF*LF	0106
	CPSI=((X(JP)-X(JC))*(X(JM)-X(JC))+(Y(JP)-Y(JC))*(Y(JM)-Y(JC))	0107
	2+(Z(JP)-Z(JC))*(Z(JM)-Z(JC)))/(DK*DL)	0108
	CALL GGMM(.O,DK,.O,DL,AM,CGDS,SGDS,SGDT,CPSI,ETA,GAM	0109
	2,Q(1,1),Q(1,2),Q(2,1),Q(2,2))	0110
	DO 98 KK=1,2	0111
	KP=IARS(KK-KG)	0112
	DO 98 LL=1,2	0113
	LP=IARS(LL-LG)	0114
	P(KP,LP)=SGN*Q(KK,LL)	0115
98	CONTINUE	0116
	GO TO 168	0117
C	K=L (SELF REACTION OF SEGMENT K)	0118
120	Q11=(.O,.O)	0119
	Q12=(.O,.O)	0120
	IF (CMM.LE.O.)GO TO 150	0121
	GD=GAM*DK	0122
	ZG=ZH/(SGDS**2)	0123
	Q11=ZG*(SGDS*CGDS-GD)/2.	0124

Fig. 8b. Subroutine SGANT

	Q12=ZG*(GD*CGDS-SGDS)/2.	0125
150	ISCK=ISC(K)	0126
	P11=(.0,.0)	0127
	P12=(.0,.0)	0128
	IF(ISCK.EQ.0)GO TO 155	0129
	IF(BM.LE.AM)GO TO 155	0130
	CALL DSHELL(AM,BM,DK,CGDS,SGDS,EP2,EP,ETA,GAM,P11,P12)	0131
155	Q11=P11+Q11	0132
	Q12=P12+Q12	0133
	CALL GGMM(.0,DK,.0,DK,AM,CGDS,SGDS,SGDS,1.	0134
	2,ETA,GAM,P11,P12,P21,P22)	0135
	Q11=P11+Q11	0136
	Q12=P12+Q12	0137
	P(1,1)=Q11	0138
	P(1,2)=Q12	0139
	P(2,1)=Q12	0140
	P(2,2)=Q11	0141
	IF(KA.NE.LA)GO TO 160	0142
	GO TO 168	0143
160	P(1,1)=-Q12	0144
	P(1,2)=-Q11	0145
	P(2,1)=-Q11	0146
	P(2,2)=-Q12	0147
168	C(MMM)=C(MMM)+FI*FJ*P(IS,JS)	0148
200	CONTINUE	0149
	DO 220 I=1,N	0150
	IJ=(I-1)*N-(I*I-1)/2+1	0151
	J1=JA(I)	0152
	IF(I2(I).EQ.IB(J1))J1=J1+NM	0153
	J2=JB(I)	0154
	IF(I2(I).EQ.IB(J2))J2=J2+NM	0155
220	C(IJ)=C(IJ)+ZLD(J1)+ZLD(J2)	0156
	RETURN	0157
	END	0158

Fig. 8c. Subroutine SGANT

In statement 168, this impedance is lumped into $C(MMM)$. The mutual impedance Z_{ij} between dipoles I and J is the sum of four segment-segment impedances.

In SGANT, segment K has endpoints KA and KB, and segment L has endpoints LA and LB. It is convenient to think of KA and KB as points 1 and 2 on segment K, and LA and LB as points 1 and 2 on L. Now we define four segment-segment impedances $P(IS,JS)$. The first subscript IS refers to the terminal point on segment K, and the second subscript JS refers to the terminal point on L. Thus $IS = 1$ or 2 if dipole I has its terminal point $I2(I)$ at KA (point 1) or KB (point 2), respectively. Similarly, $JS = 1$ or 2 if mode J has its terminal point $I2(J)$ at LA or LB. The impedances $P(IS,JS)$ are defined with the following reference directions for current flow: from point 1 toward point 2 on each segment. If dipole I has this same reference direction on segment K, we set $FI = 1$; otherwise $FI = -1$. Similarly $FJ = 1$ or -1 in accordance with the reference direction for mode J on segment L. In statement 168, $P(IS,JS)$ is multiplied by FI and FJ before its contribution is added to Z_{ij} .

Subroutine GGMM calculates the impedances $Q(KK,LL)$ which are like the $P(IS,JS)$ but have different conventions for reference directions and subscript meaning. The transformation from the Q impedances to the P impedances is accomplished in the DO LOOP ending with statement 98.

If the wire has finite conductivity, the appropriate modification is applied to the impedance matrix just above statement 150. (See Eqs. 27 through 29 in Reference 1.) The terms arising from the dielectric shell on an insulated segment are obtained from subroutine DSHELL just above statement 155. Finally, the lumped loads ZLD are added to the diagonal elements of the impedance matrix in statement 220.

The impedance matrix could be calculated in a different order as follows. Select modes I and J, calculate Z_{IJ} , and then increment I or J. Instead, SGANT selects segments K and L, calculates Z_{KL} , adds Z_{KL} to all the appropriate elements Z_{IJ} , and then increments K or L. This minimizes the calls to GGS and GGMM and presumably improves the computational efficiency.

K is a segment of test dipole I, and L is a segment of expansion mode J. When the segment numbers K and L are equal, SGANT calls GGMM to obtain the mutual impedance between two elementary electric monopoles. These monopoles are parallel and have the same length. Monopole K is positioned on the axis of the wire segment, and monopole L is on the surface of the same wire segment. Thus, the displacement is equal to the wire radius. The two monopoles are side-by-side with no stagger.

When segments K and L intersect, SGANT again calls GGMM for the mutual impedance between the two elementary monopoles. Monopole K is

situated on the axis of wire segment K, and monopole L is on the surface of wire segment L. The axes of segments K and L define a plane P, and monopole K lies in this plane. Monopole L is parallel with plane P and is displaced from it by a distance equal to the wire radius.

APPENDIX 3. Subroutine CBES

Subroutine CBES, listed in Fig. 9, calculates the quantity $B01 = J_0(z)/J_1(z)$ where z is complex and J_0 and J_1 denote the Bessel functions of order zero and one.

APPENDIX 4. Subroutine DSHELL

Subroutine DSHELL, listed in Fig. 10, calculates the mutual impedance term contributed by the dielectric insulation on the surface of a thin wire. This subroutine uses Eq. 35 of Reference 1.

APPENDIX 5. Subroutine GGS

Subroutine GGS, listed in Fig. 11, calculates the mutual impedance between two filamentary monopoles with sinusoidal current distributions. (The dipole-dipole mutual impedance in Eq. 20 of Reference 1 is the sum of four monopole-monopole mutual impedances.) The endpoints of the axial test monopole s are (X_A, Y_A, Z_A) and (X_B, Y_B, Z_B) , and the endpoints of the expansion monopole t are (X_1, Y_1, Z_1) and (X_2, Y_2, Z_2) . DS and DT denote the lengths of monopoles s and t , respectively. CAS , CBS and CGS are the direction cosines of monopole s , and CA , CB and CG are the direction cosines of monopole t .

If $INT = 0$, GGS calls GGMM for the closed-form impedance calculations. Otherwise GGS calculates the mutual impedance via Simpson's-rule integration with the following number of sample points: $IP = INT + 1$. If the monopoles are parallel with small displacement, GGS calls GGMM to avoid the difficulties of numerical integration.

For the fields of the test monopole, GGS uses Eqs. 75 and 76 of Reference 1. The current distribution on the expansion monopole is given by Eq. 74 of Reference 1. With an origin at (X_1, Y_1, Z_1) , the coordinate T measures distance along the expansion monopole. Thus T is the integration variable.

Let the coordinate s measure distance along the test monopole with origin at (X_A, Y_A, Z_A) . From any point T on monopole t , construct a line to the test monopole such that the line is perpendicular to the test monopole. SZ denotes the s coordinate of the intersection of this line with the test monopole. The length of the line is the radial coordinate ρ , and RS denotes ρ^2 . $R1$ and $R2$ are the distances from (X_A, Y_A, Z_A) and (X_B, Y_B, Z_B) to the point T . $C1$ is the current at T for the mode with terminals at (X_1, Y_1, Z_1) , and $C2$ is the current at T for the other mode with terminals at (X_2, Y_2, Z_2) . C denotes the Simpson's-rule weighting coefficient.

SUBROUTINE CBES(Z,B01)	0001
COMPLEX ARG,CC,CS,EX	0002
COMPLEX B01,Z,TERMJ,TERMN,MZ24,JN(2)	0003
DATA PI/3.14159/	0004
IF(CABS(Z).GE.12.0) GO TO 10	0005
FACTOR=0.0	0006
TERMN=(0.,0.)	0007
MZ24=-0.25*Z*Z	0008
TERMJ=(1.0,0.0)	0009
DO 1 NP=1,2	0010
N=NP-1	0011
JN(NP)=TERMJ	0012
M=0	0013
2 M=M+1	0014
TERMJ=TERMJ*MZ24/FLOAT(M*(N+M))	0015
JN(NP)=JN(NP)+TERMJ	0016
IF(NP.NE.1) GO TO 3	0017
FACTOR=FACTOR+1.0/FLOAT(M)	0018
TERMN=TERMN+TERMJ*FACTOR	0019
3 ERROR=CABS(TERMJ)	0020
IF(ERROR.GT.1.0E-10) GO TO 2	0021
1 TERMJ=0.5*Z	0022
B01=JN(1)/JN(2)	0023
RETURN	0024
10 Y=AIMAG(Z)	0025
IF(ABS(Y).GT.20.)GO TO 20	0026
ARG=(.0,1.)*Z	0027
EX=CEXP(ARG)	0028
CC=EX+1./EX	0029
CS=(.0,-1.)*(EX-1./EX)	0030
B01=(CS+CC)/(CS-CC)	0031
RETURN	0032
20 B01=(.0,-1.)	0033
IF(Y.LT.0.)B01=(.0,1.)	0034
RETURN	0035
END	0036

Fig. 9. Subroutine CBES

SUBROUTINE DSHELL(AM,BM,DK,CGDS,SGDS,EP2,EP,ETA,GAM,P11,P12)	0001
COMPLEX CGDS,SGDS,EP2,EP,ETA,GAM,P11,P12,GD,CST	0002
DATA PI/3.14159/	0003
GD=GAM*DK	0004
CST=(EP2-EP)*ETA*ALOG(BM/AM)/(4.*PI*EP2*SGDS*SGDS)	0005
P11=-CST*(GD+SGDS*CGDS)	0006
P12=CST*(GD*CGDS+SGDS)	0007
RETURN	0008
END	0009

Fig. 10. Subroutine DSHELL.

```

SUBROUTINE GGS(XA,YA,ZA,XB,YB,ZB,X1,Y1,Z1,X2,Y2,Z2,AM      0001
2,DS,CGDS,SGDS,DT,SGDT,INT,ETA,GAM,P11,P12,P21,P22)      0002
COMPLEX P11,P12,P21,P22,EJA,EJB,EJ1,EJ2,ETA,GAM,C1,C2,CST  0003
COMPLEX EGD,CGDS,SGDS,SGDT,ER1,ER2,ET1,ET2              0004
DATA FP/12.56637/                                         0005
CA=(X2-X1)/DT                                             0006
CB=(Y2-Y1)/DT                                             0007
CG=(Z2-Z1)/DT                                             0008
CAS=(XB-XA)/DS                                            0009
CBS=(YB-YA)/DS                                            0010
CGS=(ZB-ZA)/DS                                            0011
CC=CA*CAS+CB*CBS+CG*CGS                                   0012
IF(ABS(CC).GT..997)GO TO 200                              0013
20 SZ=(X1-XA)*CAS+(Y1-YA)*CBS+(Z1-ZA)*CGS                0014
IF(INT.LE.0)GO TO 300                                     0015
INS=2*(INT/2)                                             0016
IF(INS.LT.2)INS=2                                         0017
IP=INS+1                                                  0018
DELT=DT/INS                                               0019
T=.0                                                       0020
DSZ=CC*DELT                                               0021
P11=(.0,.0)                                               0022
P12=(.0,.0)                                               0023
P21=(.0,.0)                                               0024
P22=(.0,.0)                                               0025
AMS=AM*AM                                                 0026
SGN=-1.                                                    0027
DO 100 IN=1,IP                                           0028
ZZ1=SZ                                                     0029
ZZ2=SZ-DS                                                 0030
XXZ=X1+T*CA-XA-SZ*CAS                                     0031
YYZ=Y1+T*CB-YA-SZ*CBS                                     0032
ZZZ=Z1+T*CG-ZA-SZ*CGS                                     0033
RS=XXZ**2+YYZ**2+ZZZ**2                                   0034
R1=SQRT(RS+ZZ1**2)                                        0035
EJA=CEXP(-GAM*R1)                                         0036
EJ1=EJA/R1                                                0037
R2=SQRT(RS+ZZ2**2)                                        0038
EJB=CEXP(-GAM*R2)                                         0039
EJ2=EJB/R2                                                0040
ER1=EJA*SGDS+ZZ1*EJ1*CGDS-ZZ2*EJ2                      0041
ER2=-EJB*SGDS+ZZ2*EJ2*CGDS-ZZ1*EJ1                      0042
FAC=.0                                                     0043
IF(RS.GT.AMS)FAC=(CA*XXZ+CB*YYZ+CG*ZZ1)/RS              0044
ET1=CC*(EJ2-EJ1*CGDS)+FAC*ER1                            0045
ET2=CC*(EJ1-EJ2*CGDS)+FAC*ER2                            0046
C=3.+SGN                                                  0047
IF(IN.EQ.1 .OR. IN.EQ.IP)C=1.                             0048
EGD=CEXP(GAM*(DT-T))                                       0049
C1=C*(EGD-1./EGD)/2.                                       0050
EGD=CEXP(GAM*T)                                           0051
C2=C*(EGD-1./EGD)/2.                                       0052
P11=P11+ET1*C1                                           0053
P12=P12+ET1*C2                                           0054
P21=P21+ET2*C1                                           0055
P22=P22+ET2*C2                                           0056
T=T+DELT                                                  0057
SZ=SZ+DSZ                                                 0058
100 SGN=-SGN                                              0059
CST=-ETA*DELT/(3.*FP*SGDS*SGDT)                          0060
P11=CST*P11                                               0061
P12=CST*P12                                               0062

```

Fig. 11a. Subroutine GGS

	P21=CST*P21	0063
	P22=CST*P22	0064
	RETURN	0065
200	SZ1=(X1-XA)*CAS+(Y1-YA)*CBS+(Z1-ZA)*CGS	0066
	RH1=SQRT((X1-XA-SZ1*CAS)**2+(Y1-YA-SZ1*CBS)**2+(Z1-ZA-SZ1*CGS)**2)	0067
	SZ2=SZ1+DT*CC	0068
	RH2=SQRT((X2-XA-SZ2*CAS)**2+(Y2-YA-SZ2*CBS)**2+(Z2-ZA-SZ2*CGS)**2)	0069
	DDD=(RH1+RH2)/2.	0070
	IF(DDD.GT.20.*AM .AND. INT.GT.0)GO TO 20	0071
	IF(DDD.LT.AM)DDD=AM	0072
	CALL GGMM(.0,DS,SZ1,SZ2,DDD,CGDS,SGDS,SGDT,1.	0073
	2,ETA,GAM,P11,P12,P21,P22)	0074
	RETURN	0075
300	SS=SQRT(1.-CC*CC)	0076
	CAD=(CGS*CB-CBS*CG)/SS	0077
	CBD=(CAS*CG-CGS*CA)/SS	0078
	CGD=(CBS*CA-CAS*CB)/SS	0079
	DK=(X1-XA)*CAD+(Y1-YA)*CBD+(Z1-ZA)*CGD	0080
	DK=ABS(DK)	0081
	IF(DK.LT.AM)DK=AM	0082
	XZ=XA+SZ*CAS	0083
	YZ=YA+SZ*CBS	0084
	ZZ=ZA+SZ*CGS	0085
	XP1=X1-DK*CAD	0086
	YP1=Y1-DK*CBD	0087
	ZP1=Z1-DK*CGD	0088
	CAP=CBS*CGD-CGS*CBD	0089
	CBP=CGS*CAD-CAS*CGD	0090
	CGP=CAS*CBD-CBS*CAD	0091
	P1=CAP*(XP1-XZ)+CBP*(YP1-YZ)+CGP*(ZP1-ZZ)	0092
	T1=P1/SS	0093
	S1=T1*CC-SZ	0094
	CALL GGMM(S1,S1+DS,T1,T1+DT,DK,CGDS,SGDS,SGDT,CC,ETA,GAM	0095
	2,P11,P12,P21,P22)	0096
	RETURN	0097
	END	0098

Fig. 11b. Subroutine GGS

Below statement 300, GGS performs some analytic geometry in preparation for calling GGMM. The remaining part of this Appendix concerns this last part of subroutine GGS.

Let \hat{s} denote a unit vector in the direction from (X_A, Y_A, Z_A) toward (X_B, Y_B, Z_B) . Also let \hat{t} denote a unit vector from (X_1, Y_1, Z_1) toward (X_2, Y_2, Z_2) . Then $\hat{s} \cdot \hat{t} = \cos \theta = CC$ where θ is the angle formed by the axes of the two monopoles. Let monopole s lie in one plane P_s and monopole t lie in another parallel plane P_t . CAD , CBD and CGD are the direction cosines of the unit vector $\hat{d} = \hat{t} \times \hat{s} / \sin \theta$ which is perpendicular to both planes. To obtain the distance DK between the two planes, we construct a vector \underline{R}_{11} from (X_A, Y_A, Z_A) to (X_1, Y_1, Z_1) and take $DK = \underline{R}_{11} \cdot \hat{d}$.

Construct a line from (X_1, Y_1, Z_1) to the test monopole, such that the line is perpendicular to the test monopole. SZ denotes the s coordinate of the intersection of this line with the test monopole, and the cartesian coordinates of this intersection are XZ , YZ and ZZ . The direction cosines of $\hat{s} \times \hat{d}$ are CAP , CBP and CGP .

From the point (X_1, Y_1, Z_1) in plane P_t , construct a perpendicular line to the point (X_{P1}, Y_{P1}, Z_{P1}) in the plane P_s . This line is parallel with \hat{d} and has length DK . Let \underline{R} represent a vector from (XZ, YZ, ZZ) to (X_{P1}, Y_{P1}, Z_{P1}) . $P1$ denotes $\underline{R} \cdot (\hat{s} \times \hat{d})$. $S1$ and $T1$ are defined in the next Appendix.

APPENDIX 6. Subroutine GGMM

Subroutine GGMM calculates the mutual impedance between two filamentary monopoles with sinusoidal current distributions. The dipole-dipole mutual impedance in Eq. 20 of Reference 1 is the sum of four monopole-monopole mutual impedances. The monopole impedances are calculated by GGS with Simpson's rule or by GGMM with closed-form expressions in terms of exponential integrals.

To explain the input data for GGMM, reference is made to Fig. 12. Subroutine GGMM is listed in Fig. 13. If the monopoles are parallel, let the z axis be parallel with both monopoles. The coordinate origin may be selected arbitrarily. $S1$ and $S2$ denote the z coordinates of the endpoints of the test monopole, $T1$ and $T2$ are the z coordinates of the endpoints of the expansion monopole, and D is the perpendicular distance (displacement) between the monopoles. The mutual impedance of parallel monopoles is calculated in the last part of GGMM below statement 110.

For skew monopoles, let the test monopole s lie in the xy plane and the expansion monopole t in the plane $z = D$. (D is the perpendicular distance between the parallel planes.) If the monopoles are viewed along a line of sight parallel with the z axis as in Fig. 12, the extended axes of the two monopoles will appear to intersect at a point on the xy plane. Let s measure the distance along the axis of

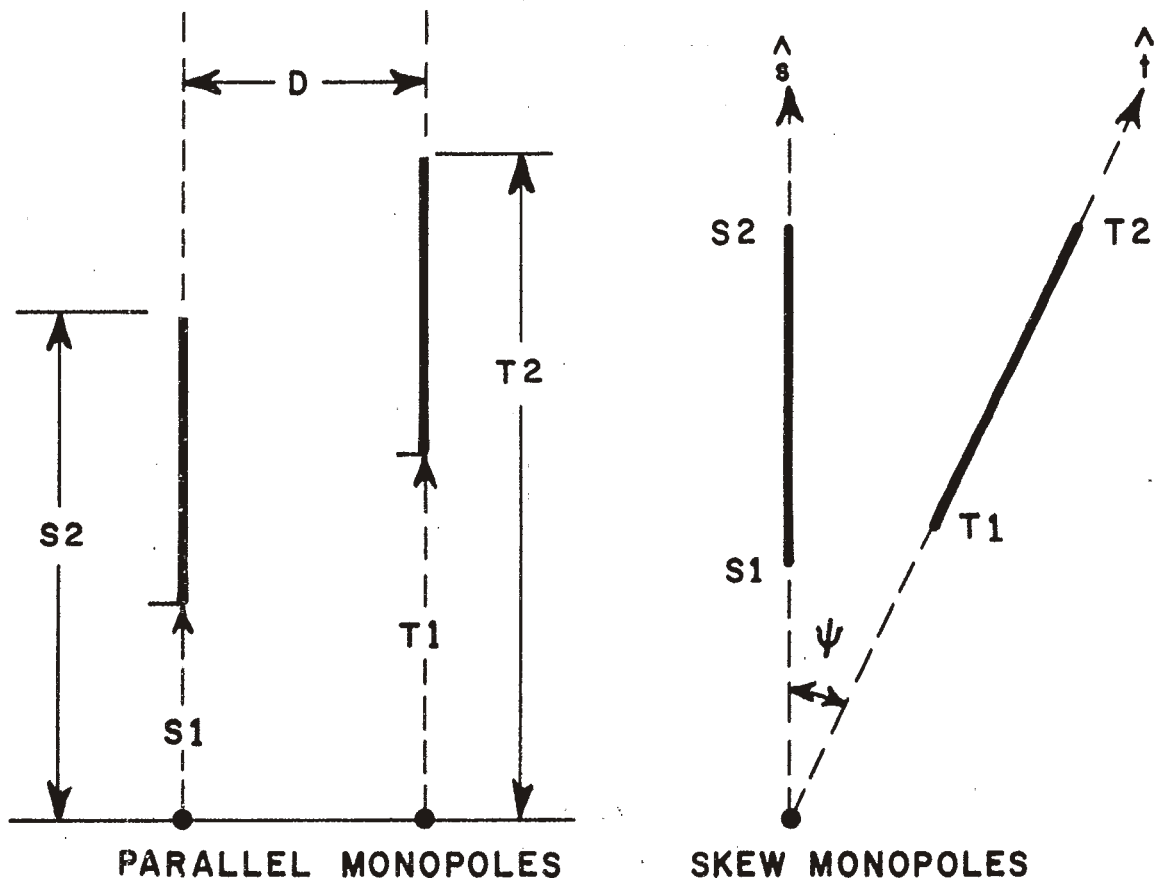


Fig. 12. Coordinates for parallel and skew monopoles in subroutine GGMM.

SUBROUTINE GGMM(S1,S2,T1,T2,D,CGDS,SGD1,SGD2,CPSI,ETA,GAM	0001
2,P11,P12,P21,P22)	0002
DOUBLE PRECISION R1,R2,DPQ,SIS,TS1,TS2,ST1,ST2,CD,BD,CPSS,SK	0003
2,TL1,TL2,TD1,TD2,SDI,DPSI,DD,ZD	0004
COMPLEX CGDS,SGDS,SGDT,SGD1,SGD2,ETA,GAM,P11,P12,P21,P22	0005
COMPLEX CST,EB,EC,EK,EL,EKL,EGZI,ES1,ES2,ET1,ET2,EXPA,EXPB	0006
COMPLEX E(2,2),F(2,2)	0007
COMPLEX EGZ(2,2),GM(2),GP(2)	0008
DATA PI/3.14159/	0009
DSQ=D*D	0010
SGDS=SGD1	0011
IF(S2.LT.S1)SGDS=-SGD1	0012
SGDT=SGD2	0013
IF(T2.LT.T1)SGDT=-SGD2	0014
IF(ABS(CPSI).GT..997)GO TO 110	0015
ES1=CEXP(GAM*S1)	0016
ES2=CEXP(GAM*S2)	0017
ET1=CEXP(GAM*T1)	0018
ET2=CEXP(GAM*T2)	0019
DD=D	0020
DPSI=CPSI	0021
TD1=T1	0022
TD2=T2	0023
CPSS=DPSI*DPSI	0024
CD=DD/DSQRT(1.0-CPSS)	0025
C=CD	0026
BD=CD*DPSI	0027
B=BD	0028
EB=CEXP(GAM*CMPLX(.0,B))	0029
EC=CEXP(GAM*CMPLX(.0,C))	0030
DO 10 K=1,2	0031
DO 10 L=1,2	0032
10 E(K,L)=(.0,.0)	0033
TS1=TD1*TD1	0034
TS2=TD2*TD2	0035
DPQ=DD*DD	0036
SI=S1	0037
DO 100 I=1,2	0038
FI=(-1)**I	0039
SDI=SI	0040
SIS=SDI*SDI	0041
ST1=2.*SDI*TD1*DPSI	0042
ST2=2.*SDI*TD2*DPSI	0043
R1=DSQRT(DPQ+SIS+TS1-ST1)	0044
R2=DSQRT(DPQ+SIS+TS2-ST2)	0045
EK=EB	0046
DO 50 K=1,2	0047
FK=(-1)**K	0048
SK=FK*SDI	0049
EL=EC	0050
DO 40 L=1,2	0051
FL=(-1)**L	0052
EKL=EK*EL	0053
XX=FK*BD+FL*CD	0054
TL1=FL*TD1	0055
TL2=FL*TD2	0056
RR1=R1+SK+TL1	0057
RR2=R2+SK+TL2	0058
CALL EXPJ(GAM*CMPLX(RR1,-XX),GAM*CMPLX(RR2,-XX),EXPA)	0059
CALL EXPJ(GAM*CMPLX(RR1,XX),GAM*CMPLX(RR2,XX),EXPB)	0060
E(K,L)=E(K,L)+FI*(EXPA*EKL+EXPB/EKL)	0061
40 EL=1./EC	0062

Fig. 13a. Subroutine GGMM

50	EK=1./EB	0063
	ZD=SDI*DPSI	0064
	ZC=ZD	0065
	EGZI=CEXP(GAM*ZC)	0066
	RR1=R1+ZD-TD1	0067
	RR2=R2+ZD-TD2	0068
	CALL EXPJ(GAM*RR1,GAM*RR2,EXPB)	0069
	RR1=R1-ZD+TD1	0070
	RR2=R2-ZD+TD2	0071
	CALL EXPJ(GAM*RR1,GAM*RR2,EXPA)	0072
	F(I,1)=2.*SGDS*EXPA/EGZI	0073
	F(I,2)=2.*SGDS*EXPB*EGZI	0074
100	S1=S2	0075
	CST=ETA/(16.*PI*SGDS*SGDT)	0076
	P11=CST*((F(1,1)+E(2,2)*ES2-E(1,2)/ES2)*ET2	0077
A	+(-F(1,2)-E(2,1)*ES2+E(1,1)/ES2)/ET2)	0078
	P12=CST*((-F(1,1)-E(2,2)*ES2+E(1,2)/ES2)*ET1	0079
B	+(F(1,2)+E(2,1)*ES2-E(1,1)/ES2)/ET1)	0080
	P21=CST*((-F(2,1)-E(2,2)*ES1+E(1,2)/ES1)*ET2	0081
C	+(F(2,2)+E(2,1)*ES1-E(1,1)/ES1)/ET2)	0082
	P22=CST*((F(2,1)+E(2,2)*ES1-E(1,2)/ES1)*ET1	0083
D	+(-F(2,2)-E(2,1)*ES1+E(1,1)/ES1)/ET1)	0084
	RETURN	0085
110	IF(CPSI.LT.0.)GO TO 120	0086
	TA=T1	0087
	TB=T2	0088
	GO TO 130	0089
120	TA=-T1	0090
	TB=-T2	0091
	SGDT=-SGDT	0092
130	S1=S1	0093
	DO 150 I=1,2	0094
	TJ=TA	0095
	DO 140 J=1,2	0096
	ZIJ=TJ-SI	0097
	R=SQRT(DSQ+ZIJ*ZIJ)	0098
	W=R+ZIJ	0099
	IF(ZIJ.LT.0.)W=DSQ/(R-ZIJ)	0100
	V=R-ZIJ	0101
	IF(ZIJ.GT.0.)V=DSQ/(R+ZIJ)	0102
	IF(J.EQ.1)V1=V	0103
	IF(J.EQ.1)W1=W	0104
	EGZ(I,J)=CEXP(GAM*ZIJ)	0105
140	TJ=TB	0106
	CALL EXPJ(GAM*V1,GAM*V,GP(I))	0107
	CALL EXPJ(GAM*W1,GAM*W,GM(I))	0108
150	S1=S2	0109
	CST=-ETA/(8.*PI*SGDS*SGDT)	0110
	P11=CST*(GM(2)*EGZ(2,2)+GP(2)/EGZ(2,2)	0111
	2-CGDS*(GM(1)*EGZ(1,2)+GP(1)/EGZ(1,2)))	0112
	P12=CST*(-GM(2)*EGZ(2,1)-GP(2)/EGZ(2,1)	0113
	2+CGDS*(GM(1)*EGZ(1,1)+GP(1)/EGZ(1,1)))	0114
	P21=CST*(GM(1)*EGZ(1,2)+GP(1)/EGZ(1,2)	0115
	2-CGDS*(GM(2)*EGZ(2,2)+GP(2)/EGZ(2,2)))	0116
	P22=CST*(-GM(1)*EGZ(1,1)-GP(1)/EGZ(1,1)	0117
	2+CGDS*(GM(2)*EGZ(2,1)+GP(2)/EGZ(2,1)))	0118
	RETURN	0119
	END	0120

Fig. 13b. Subroutine GGMM

the test monopole with origin at the apparent intersection. S1 and S2 denote the s coordinates of the endpoints of the test monopole. Similarly, let t measure distance along the axis of the expansion monopole with origin at the apparent intersection. T1 and T2 denote the t coordinates of the endpoints of the expansion monopole. Let \hat{s} and \hat{t} be unit vectors parallel with the positive s and t axes, respectively. Then $\text{CPSI} = \hat{s} \cdot \hat{t} = \cos \psi$. The monopole lengths are d_s and d_t , and the remaining input data are defined as follows:

CGDS	$\cosh \gamma d_s$
SGD1	$\sinh \gamma d_s$
SGD2	$\sinh \gamma d_t$

GGMM calls EXPJ for the exponential integrals.

The output data from GGMM are the impedances P11, P12, P21, and P22. In defining these impedances, the reference direction is from S1 to S2 for the current on monopole s, and from T1 to T2 for the current on monopole t. In the impedance P_{ij} , the first subscript is 1 or 2 if the test dipole has terminals at S1 or S2 on monopole s. The second subscript is 1 or 2 if the expansion dipole has terminals at T1 or T2 on monopole t. The endpoint coordinates S1, S2, T1 and T2 may be positive or negative. The monopole lengths d_s and d_t are assumed positive in defining the input data CGDS, SGD1 and SGD2.

For parallel monopoles, $\text{CPSI} = 1$ or -1 . S1, S2, T1 and T2 are cartesian coordinates for parallel monopoles and spherical coordinates for skew monopoles. For skew monopoles, the radial coordinates S1, S2, T1 and T2 tend to infinity as the angle ψ tends to zero or π . Therefore, if the monopoles are within 4.5° of being parallel, they are approximated by parallel dipoles.

APPENDIX 7. Subroutine EXPJ

Subroutine EXPJ, listed in Fig. 14, evaluates the exponential integral defined as follows:

$$(2) \quad W_{12} = \int_{V_1}^{V_2} \frac{e^{-v}}{v} dv = E_1(V_1) - E_1(V_2) + j 2n\pi$$

where the integration path is the straight line from V1 to V2 on the complex v plane and

$$(3) \quad E_1(z) = \int_z^\infty \frac{e^{-t}}{t} dt$$

	SUBROUTINE EXPJ(V1,V2,W12)	0001
	COMPLEX EC,E15,S,T,UC,VC,V1,V2,W12,Z	0002
	DIMENSION V(21),W(21),D(16),E(16)	0003
	DATA V/ 0.22284667E 00,	0004
	20.11889321E 01,0.29927363E 01,0.57751436E 01,0.98374674E 01,	0005
	20.15982874E 02,0.93307812E-01,0.49269174E 00,0.12155954E 01,	0006
	20.22699495E 01,0.36676227E 01,0.54253366E 01,0.75659162E 01,	0007
	20.10120228E 02,0.13130282E 02,0.16654408E 02,0.20776479E 02,	0008
	20.25623894E 02,0.31407519E 02,0.38530683E 02,0.48026086E 02/	0009
	DATA W/ 0.45896460E 00,	0010
	20.41700083E 00,0.11337338E 00,0.10399197E-01,0.26101720E-03,	0011
	20.89854791E-06,0.21823487E 00,0.34221017E 00,0.26302758E 00,	0012
	20.12642582E 00,0.40206865E-01,0.85638778E-02,0.12124361E-02,	0013
	20.11167440E-03,0.64599267E-05,0.22263169E-06,0.42274304E-08,	0014
	20.39218973E-10,0.14565152E-12,0.14830270E-15,0.16005949E-19/	0015
	DATA D/ 0.22495842E 02,	0016
	2 0.74411568E 02,-0.41431576E 03,-0.78754339E 02, 0.11254744E 02,	0017
	2 0.16021761E 03,-0.23862195E 03,-0.50094687E 03,-0.68487854E 02,	0018
	2 0.12254778E 02,-0.10161976E 02,-0.47219591E 01, 0.79729681E 01,	0019
	2-0.21069574E 02, 0.22046490E 01, 0.89728244E 01/	0020
	DATA E/ 0.21103107E 02,	0021
	2-0.37959787E 03,-0.97489220E 02, 0.12900672E 03, 0.17949226E 02,	0022
	2-0.12910931E 03,-0.55705574E 03, 0.13524801E 02, 0.14696721E 03,	0023
	2 0.17949528E 02,-0.32981014E 00, 0.31028836E 02, 0.81657657E 01,	0024
	2 0.22236961E 02, 0.39124892E 02, 0.81636799E 01/	0025
	Z=V1	0026
	DQ 100 JIM=1,2	0027
	X=REAL(Z)	0028
	Y=AIMAG(Z)	0029
	E15=(.0,.0)	0030
	AB=CABS(Z)	0031
	IF(AB.EQ.0.)GO TO 90	0032
	IF(X.GE.0. .AND. AB.GT.10.)GO TO 80	0033
	YA=ABS(Y)	0034
	IF(X.LE.0. .AND. YA.GT.10.)GO TO 80	0035
	IF(YA-X.GE.17.5.OR.YA.GE.6.5.OR.X+YA.GE.5.5 .OR.X.GE.3.)GO TO 20	0036
	IF(X.LE.-9.)GO TO 40	0037
	IF(YA-X.GE.2.5)GO TO 50	0038
	IF(X+YA.GE.1.5)GO TO 30	0039
10	N=6.+3.*AB	0040
	E15=1./(N-1.)-Z/N**2	0041
15	N=N-1	0042
	E15=1./(N-1.)-Z*E15/N	0043
	IF(N.GE.3)GO TO 15	0044
	E15=Z*E15-CMPLX(.577216+ALOG(AB),ATAN2(Y,X))	0045
	GO TO 90	0046
20	J1=1	0047
	J2=6	0048
	GO TO 31	0049
30	J1=7	0050
	J2=21	0051
31	S=(.0,.0)	0052
	YS=Y*Y	0053
	DO 32 I=J1,J2	0054
	XI=V(I)+X	0055
	CF=W(I)/(XI*XI+YS)	0056
32	S=S+CMPLX(XI*CF,-YA*CF)	0057
	GO TO 54	0058
40	T3=X*X-Y*Y	0059
	T4=2.*X*YA	0060
	T5=X*T3-YA*T4	0061
	T6=X*T4+YA*T3	0062

Fig. 14a. Subroutine EXPJ

```

UC=CMPLX(D(11)+D(12)*X+D(13)*T3+T5-E(12)*YA-E(13)*T4,      0063
2      E(11)+E(12)*X+E(13)*T3+T6+D(12)*YA+D(13)*T4)         0064
VC=CMPLX(D(14)+D(15)*X+D(16)*T3+T5-E(15)*YA-E(16)*T4,      0065
2      E(14)+E(15)*X+E(16)*T3+T6+D(15)*YA+D(16)*T4)         0066
GO TO 52                                                       0067
50 T3=X*X-Y*Y                                                 0068
T4=2.*X*YA                                                    0069
T5=X*T3-YA*T4                                                 0070
T6=X*T4+YA*T3                                                 0071
T7=X*T5-YA*T6                                                 0072
T8=X*T6+YA*T5                                                 0073
T9=X*T7-YA*T8                                                 0074
T10=X*T8+YA*T7                                                0075
UC=CMPLX(D(1)+D(2)*X+D(3)*T3+D(4)*T5+D(5)*T7+T9-(E(2)*YA+E(3)*T4
2+E(4)*T6+E(5)*T8),E(1)+E(2)*X+E(3)*T3+E(4)*T5+E(5)*T7+T10+
3(D(2)*YA+D(3)*T4+D(4)*T6+D(5)*T8))                          0076
VC=CMPLX(D(6)+D(7)*X+D(8)*T3+D(9)*T5+D(10)*T7+T9-(E(7)*YA+E(8)*T4
2+E(9)*T6+E(10)*T8),E(6)+E(7)*X+E(8)*T3+E(9)*T5+E(10)*T7+T10+
3(D(7)*YA+D(8)*T4+D(9)*T6+D(10)*T8))                          0077
52 EC=UC/VC                                                    0078
S=EC/CMPLX(X,YA)                                              0079
54 EX=EXP(-X)                                                  0080
T=EX*CMPLX(COS(YA),-SIN(YA))                                 0081
E15=S*T                                                       0082
56 IF(Y.LT.0.)E15=CONJG(E15)                                  0083
GO TO 90                                                       0084
80 E15=.409319/(Z+.193044)+.421831/(Z+1.02666)+.147126/(Z+2.56788)+
2.206335E-1/(Z+4.90035)+.107401E-2/(Z+8.18215)+.158654E-4/(Z+
312.7342)+.317031E-7/(Z+19.3957)                             0085
E15=E15*CEXP(-Z)                                             0086
90 IF(JIM.EQ.1)W12=E15                                        0087
100 Z=V2                                                       0088
Z=V2/V1                                                       0089
TH=ATAN2(AIMAG(Z),REAL(Z))-ATAN2(AIMAG(V2),REAL(V2))         0090
2+ATAN2(AIMAG(V1),REAL(V1))                                  0091
AB=ABS(TH)                                                    0092
IF(AB.LT.1.)TH=.0                                            0093
IF(TH.GT.1.)TH=6.2831853                                     0094
IF(TH.LT.-1.)TH=-6.2831853                                  0095
W12=W12-E15+CMPLX(.0,TH)                                     0096
RETURN                                                         0097
END                                                            0098

```

Fig. 14b. Subroutine EXPJ

The exponential integral $E_1(z)$ is defined in Reference 3. To generate W12, subroutine EXPJ calculates $E_1(V1)$, subtracts $E_1(V2)$ and adds $j2n\pi$. The term $j2n\pi$ is determined by the requirement that W12 vanish in the limit as $V1$ approaches $V2$. The integer n may assume values of -1 , 0 or $+1$. If the integration path does not cross the negative real axis in the v plane, n is zero. The term $j2n\pi$ is calculated below statement 100.

APPENDIX 8. Subroutine GANT1

Subroutine GANT1, listed in Fig. 15, considers the wire structure as an antenna. In the input data, $VG(J)$ is the voltage of a generator at point $IA(J)$ of segment J . $VG(JJ)$ is the voltage of a generator at point $IB(J)$ of segment J . The DO LOOP ending with statement 50 uses the delta-gap model to determine the excitation voltages $CJ(I)$ for all the dipole modes. These are also stored temporarily in $CG(I)$. Then subroutine SQROT is called to obtain a solution of the simultaneous linear equations. SQROT stores the solution (the loop currents) in $CJ(I)$.

In the DO LOOP ending at statement 80, the complex power input is calculated and stored in $Y11$. GG denotes the time-average power input and is the real part of $Y11$. If the antenna has only one voltage generator (with unit voltage and zero phase angle), then $Y11$ also denotes the antenna admittance and $Z11$ is the antenna impedance at that port.

Subroutine RITE is called to make the transformation from the loop currents $CJ(I)$ to the branch currents $CG(J)$. If IWR is a positive integer, RITE will write out the list of branch currents.

Finally, GANT1 calculates the radiation efficiency EFF . PIN denotes the time-average power input. Subroutine GDISS is called to obtain the time-average power dissipated. $DISS$ is the total power dissipated in the lumped loads and the imperfectly-conducting wire. $PRAD$ is the time-average power radiated, defined by the difference between PIN and $DISS$. If the antenna has perfect conductivity and purely reactive loads, the radiation efficiency is considered to be 100 per cent.

APPENDIX 9. Subroutine SQROT

Subroutine SQROT is listed in Fig. 16. This subroutine considers the matrix equation $ZI = V$ which represents a system of simultaneous linear equations. If the square matrix Z is symmetric, SQROT is useful for obtaining the solution I with V given. NEQ denotes the number of simultaneous equations and the size of the matrix Z .

On entry to SQROT, S is the excitation column V . On exit, the solution I is stored in S . Let $Z(I,J)$ denote the symmetric square

	SUBROUTINE GANT1(IA,IB,INM,IWR,I1,I2,I3,I12,JA,JB,MD,N,ND,NM,AM	0001
	2,C,CJ,CG,CMM,D,EFF,GAM,GG,CGD,SGD,VG,Y11,Z11,ZLD,ZS)	0002
	COMPLEX C(1),CJ(1),CGD(1),SGD(1),VG(1),ZLD(1),Y11,Z11,ZS,GAM,CG(1)	0003
	DIMENSION D(1),IA(1),IB(1),JA(1),JB(1)	0004
	DIMENSION I1(1),I2(1),I3(1),MD(INM,4),ND(1)	0005
2	FORMAT(1X,1I5,8F10.2)	0006
5	FORMAT(1H0)	0007
	DO 50 I=1,N	0008
	CJ(I)=(.0,.0)	0009
	K=JA(I)	0010
	DO 40 KK=1,2	0011
	KA=IA(K)	0012
	KB=IB(K)	0013
	JJ=K	0014
	FI=1.	0015
	IF(KB.EQ.I2(I))GO TO 36	0016
	IF(KB.EQ.I1(I))FI=-1.	0017
	CJ(I)=CJ(I)+FI*VG(JJ)	0018
	GO TO 40	0019
36	IF(KA.EQ.I3(I))FI=-1.	0020
	JJ=K+NM	0021
	CJ(I)=CJ(I)+FI*VG(JJ)	0022
40	K=JB(I)	0023
50	CONTINUE	0024
	DO 55 I=1,N	0025
55	CG(I)=CJ(I)	0026
	CALL SQROT(C,CJ,0,I12,N)	0027
	I12=2	0028
	Y11=(.0,.0)	0029
	DO 80 I=1,N	0030
80	Y11=Y11+CJ(I)*CONJG(CG(I))	0031
	CALL RITE(IA,IB,INM,IWR,I1,I2,I3,MD,ND,NM,CJ,CG)	0032
	GG=REAL(Y11)	0033
	Z11=1./Y11	0034
	PIN=GG	0035
	CALL GDISS(AM,CG,CMM,D,DISS,GAM,NM,SGD,ZLD,ZS)	0036
	PRAD=PIN-DISS	0037
	EFF=100.*PRAD/PIN	0038
	RETURN	0039
	END	0040

Fig. 15. Subroutine GANT1

	SUBROUTINE SQROT(C,S,IWR,I12,NEQ)	0001
	COMPLEX C(1),S(1),SS	0002
2	FORMAT(1X,1I5,1F10.3,1F15.7,1F10.0,2F15.6)	0003
3	FORMAT(1H0)	0004
	N=NEQ	0005
	IF (I12.EQ.2)GO TO 20	0006
	C(1)=CSQRT(C(1))	0007
	DO 4 K=2,N	0008
4	C(K)=C(K)/C(1)	0009
	DO 10 I=2,N	0010
	IM0=I-1	0011
	IPO=I+1	0012
	ID=(I-1)*N-(I*I-I)/2	0013
	II=ID+I	0014
	DO 5 L=1,IM0	0015
	LI=(L-1)*N-(L*L-L)/2+I	0016
5	C(II)=C(II)-C(LI)*C(LI)	0017
	C(II)=CSQRT(C(II))	0018
	IF (IPO.GT.N)GO TO 10	0019
	DO 8 J=IPO,N	0020
	IJ=IO+J	0021
	DO 6 M=1,IM0	0022
	MD=(M-1)*N-(M*M-M)/2	0023
	MI=MD+I	0024
	MJ=MD+J	0025
6	C(IJ)=C(IJ)-C(MJ)*C(MI)	0026
8	C(IJ)=C(IJ)/C(II)	0027
10	CONTINUE	0028
20	S(1)=S(1)/C(1)	0029
	DO 30 I=2,N	0030
	IM0=I-1	0031
	DO 25 L=1,IM0	0032
	LI=(L-1)*N-(L*L-L)/2+I	0033
25	S(I)=S(I)-C(LI)*S(L)	0034
	II=(I-1)*N-(I*I-I)/2+I	0035
30	S(I)=S(I)/C(II)	0036
	NN=(N+1)*N/2	0037
	S(N)=S(N)/C(NN)	0038
	NM0=N-1	0039
	DO 40 I=1,NM0	0040
	K=N-I	0041
	KPO=K+1	0042
	KD=(K-1)*N-(K*K-K)/2	0043
	DO 35 L=KPO,N	0044
	KL=KD+L	0045
35	S(K)=S(K)-C(KL)*S(L)	0046
	KK=KD+K	0047
40	S(K)=S(K)/C(KK)	0048
	IF (IWR.LE.0) GO TO 100	0049
	CNOR=.0	0050
	DO 50 I=1,N	0051
	SA=CABS(S(I))	0052
50	IF (SA.GT.CNOR)CNOR=SA	0053
	IF (CNOR.LE.0.)CNOR=1.	0054
	DO 60 I=1,N	0055
	SS=S(I)	0056
	SA=CABS(SS)	0057
	SNOR=SA/CNOR	0058
	PH=.0	0059
	IF (SA.GT.0.)PH=57.29578*ATAN2(AIMAG(SS),REAL(SS))	0060
60	WRITE (6,2)I,SNOR,SA,PH,SS	0061
	WRITE (6,3)	0062
100	RETURN	0063
	END	0064

Fig. 16. Subroutine SQROT

matrix. On entry to SQROT, the upper-right triangular portion of $Z(I,J)$ is stored by rows in $C(K)$ with

$$(4) \quad K = (I - 1) * NEQ - (I * I - I) / 2 + J$$

If $I12 = 1$, SQROT will transform the symmetric matrix into the auxiliary matrix (implicit inverse), store the result in $C(K)$ and use the auxiliary matrix to solve the simultaneous equations. If $I12 = 2$, this indicates that $C(K)$ already contains the auxiliary matrix.

The transformation from the symmetric matrix to the auxiliary matrix is programmed above statement 10, and the solution of the simultaneous equations is programmed in statements 20 to 40. If IWR is positive, the program below statement 40 will write the solution.

SQROT uses the square root method described in Reference 4. The original symmetric matrix Z and the upper triangular auxiliary matrix A are related by

$$(5) \quad Z = A' A$$

where A' is the transpose of A .

In the thin-wire application, SQROT must be called with $I12 = 1$ before it is called with $I12 = 2$. With a large matrix, the execution time in SQROT is much smaller with $I12 = 2$ than with $I12 = 1$.

APPENDIX 10. Subroutine RITE

Subroutine RITE is listed in Fig. 17. Given the list of loop currents $CJ(I)$, this subroutine generates a list of branch currents $CG(J)$. $CG(J)$ and $CG(JJ)$ denote the currents at $IA(J)$ and $IB(J)$, respectively, on the wire segment J , where $JJ = J + NM$. If IWR is a positive integer, the program below statement 110 writes a list of the branch currents. The symbols in this list are defined as follows:

K	the segment number
ACJ	normalized current magnitude at $IA(K)$
BCJ	normalized current magnitude at $IB(K)$
PA	phase of current at $IA(K)$
PB	phase of current at $IB(K)$
CJA	complex current at $IA(K)$
CJB	complex current at $IB(K)$

The phase angles PA and PB are in degrees. Even if IWR is negative, RITE generates the branch-current list for use in subroutine $GDISS$.

	SUBROUTINE RITE(IA,IB,INM,IWR,I1,I2,I3,MD,ND,NM,CJ,CG)	0001
	COMPLEX CJ(1),CG(1),CJA,CJB	0002
	DIMENSION IA(1),IB(1),I1(1),I2(1),I3(1),MD(INM,4),ND(1)	0003
2	FORMAT(1X,1I5,2F10.3,2F10.0,4F15.6)	0004
5	FORMAT(1H0)	0005
	AMAX=.0	0006
	DO 100 K=1,NM	0007
	KA=IA(K)	0008
	KB=IB(K)	0009
	CJA=(.0,.0)	0010
	CJB=(.0,.0)	0011
	NDK=ND(K)	0012
	DO 40 II=1,NDK	0013
	I=MD(K,II)	0014
	FI=1.	0015
	IF(KB.EQ.I2(II))GO TO 36	0016
	IF(KB.EQ.I1(II))FI=-1.	0017
	CJA=CJA+FI*CJ(I)	0018
	GO TO 40	0019
36	IF(KA.EQ.I3(II))FI=-1.	0020
	CJB=CJB+FI*CJ(I)	0021
40	CONTINUE	0022
	CG(K)=CJA	0023
	KK=K+NM	0024
	CG(KK)=CJB	0025
	ACJ=CABS(CJA)	0026
	BCJ=CABS(CJB)	0027
	IF(ACJ.GT.AMAX)AMAX=ACJ	0028
	IF(BCJ.GT.AMAX)AMAX=BCJ	0029
100	CONTINUE	0030
	IF(IWR.GT.0)GO TO 110	0031
	RETURN	0032
110	IF(AMAX.LE.0.)AMAX=1.	0033
	DO 200 K=1,NM	0034
	CJA=CG(K)	0035
	KK=K+NM	0036
	CJB=CG(KK)	0037
	ACJ=CABS(CJA)/AMAX	0038
	BCJ=CABS(CJB)/AMAX	0039
	PA=57.29578*ATAN2(AIMAG(CJA),REAL(CJA))	0040
	PB=57.29578*ATAN2(AIMAG(CJB),REAL(CJB))	0041
200	WRITE(6,2)K,ACJ,BCJ,PA,PB,CJA,CJB	0042
	WRITE(6,5)	0043
	RETURN	0044
	END	0045

Fig. 17. Subroutine RITE

APPENDIX 11. Subroutine GDISS

Subroutine GDISS is listed in Fig. 18. This subroutine uses Eq. 50 of Reference 1 to calculate the time-average power dissipated in the imperfectly conducting wire. This is accomplished in the DO LOOP terminating at statement 100. The power dissipated in the lumped loads is calculated in the DO LOOP terminating with statement 140. DISS denotes the time-average power dissipated in the wire and the loads.

APPENDIX 12. Subroutine GNFLD

Subroutine GNFLD, listed in Fig. 19, inputs the loop currents $CJ(I)$, calls GNF for the near-zone field of each wire segment, and sums over all the segments to obtain the near-zone field of the wire antenna or the near-zone scattered field of the wire scatterer. EX, EY and EZ denote the cartesian components of this field at the observation point (XP,YP,ZP). This calculated field does not include the incident fields of the magnetic frills or loops associated with generators on the antenna. It also does not include the radiation from the polarization currents in the dielectric insulation.

This subroutine could be simplified and speeded by inputting the branch currents $CG(J)$ instead of the loop currents $CJ(I)$. However, this would increase the storage requirements because the far-field subroutine GFFLD would have to store the branch currents induced by the phi-polarized and theta-polarized incident waves.

APPENDIX 13. Subroutine GNF

Subroutine GNF, listed in Fig. 20, uses Eqs. 75 and 76 of Reference 1 to calculate the near-zone electric field of a sinusoidal electric monopole with endpoints at (XA,YA,ZA) and (XB,YB,ZB) . The observation point is at (X,Y,Z) . EX1, EY1 and EZ1 are the components of the field generated by the mode with unit current at (XA,YA,ZA) . EX2, EY2 and EZ2 denote the field generated by the mode with unit current at (XB,YB,ZB) . GNF is similar to GGS, and Appendix 5 defines many of the symbols used in both subroutines.

APPENDIX 14. Subroutine GFFLD

The far-field subroutine GFFLD, listed in Fig. 21, is discussed in section II. In antenna gain calculations with $INC = 0$, the loop currents $CJ(I)$ are employed by GFFLD to calculate the far-zone field. The field of each segment is obtained by calling GFF, and a summation over all the segments yields the field of the antenna.

In a bistatic scattering situation with $INC = 2$, the input data include the loop currents EP and ET induced by phi-polarized and theta-polarized incident waves. These currents were calculated by GFFLD in a

SUBROUTINE GDISS(AM,CG,CMM,D,DISS,GAM,NM,SGD,ZLD,ZS)	0001
COMPLEX CG(1),SGD(1),ZLD(1),CJA,CJB,GAM,ZS	0002
DIMENSION D(1)	0003
DATA PI/3.14159/	0004
DISS=.0	0005
IF (CMM.LE.0.)GO TO 120	0006
ALPH=REAL(GAM)	0007
BETA=AIMAG(GAM)	0008
RH=REAL(ZS)/(4.*PI*AM)	0009
DO 100 K=1,NM	0010
DK=D(K)	0011
DEN=CABS(SGD(K))**2	0012
EAD=EXP(ALPH*DK)	0013
CAD=(EAD+1./EAD)/2.	0014
CBD=COS(BETA*DK)	0015
SAD=DK	0016
IF (ALPH.NE.0.)SAD=(EAD-1./EAD)/(2.*ALPH)	0017
SBD=DK	0018
IF (BETA.NE.0.)SBD=SIN(BETA*DK)/BETA	0019
FA=RH*(SAD*CAD-SBD*CBD)/DEN	0020
FB=2.*RH*(CAD*SBD-SAD*CBD)/DEN	0021
CJA=CG(K)	0022
L=K+NM	0023
CJB=CG(L)	0024
100 DISS=DISS+FA*(CABS(CJA)**2+CABS(CJB)**2)	0025
2+FB*(REAL(CJA)*REAL(CJB)+AIMAG(CJA)*AIMAG(CJB))	0026
120 DO 140 J=1,NM	0027
K=J+NM	0028
140 DISS=DISS+REAL(ZLD(J))*(CABS(CG(J))**2)	0029
2+REAL(ZLD(K))*(CABS(CG(K))**2)	0030
RETURN	0031
END	0032

Fig. 18. Subroutine GDISS

SUBROUTINE GNFLD(IA,IB,INM,I1,I2,I3,MD,N,ND,NM,AM,CGD,SGD,ETA,GAM	0001
2,CJ,D,X,Y,Z,XP,YP,ZP,EX,EY,EZ)	0002
COMPLEX EX,EY,EZ,EX1,EY1,EZ1,EX2,EY2,EZ2,ETA,GAM	0003
COMPLEX CJ(1),CGD(1),SGD(1)	0004
DIMENSION IA(1),IB(1),I1(1),I2(1),I3(1),D(1),X(1),Y(1),Z(1)	0005
DIMENSION MD(INM,4),ND(1)	0006
DATA PI,TP/3.14159,6.28318/	0007
EX=(.0,.0)	0008
EY=(.0,.0)	0009
EZ=(.0,.0)	0010
DO 140 K=1,NM	0011
KA=IA(K)	0012
KB=IB(K)	0013
CALL GNF(X(KA),Y(KA),Z(KA),X(KB),Y(KB),Z(KB),XP,YP,ZP,AM,D(K)	0014
2,CGD(K),SGD(K),ETA,GAM,EX1,EY1,EZ1,EX2,EY2,EZ2)	0015
NDK=ND(K)	0016
DO 140 I1=1,NDK	0017
I=MD(K,I1)	0018
FI=1.	0019
IF(KB.EQ.I2(I1))GO TO 136	0020
IF(KB.EQ.I1(I1))FI=-1.	0021
EX=EX+FI*EX1*CJ(I)	0022
EY=EY+FI*EY1*CJ(I)	0023
EZ=EZ+FI*EZ1*CJ(I)	0024
GO TO 140	0025
136 IF(KA.EQ.I3(I1))FI=-1.	0026
EX=EX+FI*EX2*CJ(I)	0027
EY=EY+FI*EY2*CJ(I)	0028
EZ=EZ+FI*EZ2*CJ(I)	0029
140 CONTINUE	0030
RETURN	0031
END	0032

Fig. 19. Subroutine GNFLD

SUBROUTINE GNF (XA, YA, ZA, XB, YB, ZB, X, Y, Z, AM, DS, CGDS, SGDS, ETA, GAM	0001
2, EX1, LY1, EZ1, EX2, EY2, EZ2)	0002
COMPLEX EJA, EJB, EJ1, EJ2, ER1, ER2, ES1, ES2, SGDS, GAM, CST, CGDS, ETA	0003
COMPLEX EX1, EY1, EZ1, EX2, EY2, EZ2	0004
DATA PI/3.14159/	0005
CAS=(XB-XA)/DS	0006
CBS=(YB-YA)/DS	0007
CGS=(ZB-ZA)/DS	0008
SZ=(X-XA)*CAS+(Y-YA)*CBS+(Z-ZA)*CGS	0009
ZZ1=SZ	0010
ZZ2=SZ-DS	0011
XXZ=X-XA-SZ*CAS	0012
YYZ=Y-YA-SZ*CBS	0013
ZZZ=Z-ZA-SZ*CGS	0014
RS=XXZ**2+YYZ**2+ZZZ**2	0015
R1=SQRT(RS+ZZ1**2)	0016
EJA=CEXP(-GAM*R1)	0017
EJ1=EJA/R1	0018
R2=SQRT(RS+ZZ2**2)	0019
EJB=CEXP(-GAM*R2)	0020
EJ2=EJB/R2	0021
ES1=EJ2-EJ1*CGDS	0022
ES2=EJ1-EJ2*CGDS	0023
ER1=(.0,.0)	0024
ER2=(.0,.0)	0025
AMS=AM*AM	0026
IF (RS.LT.AMS)GO TO 80	0027
CTH1=ZZ1/R1	0028
CTH2=ZZ2/R2	0029
ER1=(EJA*SGDS+EJA*CGDS*CTH1-EJB*CTH2)/RS	0030
ER2=(-EJB*SGDS+EJB*CGDS*CTH2-EJA*CTH1)/RS	0031
80 CST=ETA/(4.*PI*SGDS)	0032
EX1=CST*(ES1*CAS+ER1*XXZ)	0033
EY1=CST*(ES1*CBS+ER1*YYZ)	0034
EZ1=CST*(ES1*CGS+ER1*ZZZ)	0035
EX2=CST*(ES2*CAS+ER2*XXZ)	0036
EY2=CST*(ES2*CBS+ER2*YYZ)	0037
EZ2=CST*(ES2*CGS+ER2*ZZZ)	0038
RETURN	0039
END	0040

Fig. 20. Subroutine GNF

	SUBROUTINE GFFLD(IA,IB,INC,INM,IWR,I1,I2,I3,I12,MD,N,ND,NM,AM	0001
	2,ACSP,ACST,C,CGD,CG,CJ,CMM,D,ECSP,ECST,EP,ET,EPP,ETT,EPPS,EPTS	0002
	3,ETPS,ETTS,GG,GPP,GTT,PH,SGD,SCSP,SCST,SPPM,SPTM,STPM,STTM,TH	0003
	4,X,Y,Z,ZLD,ZS,ETA,GAM)	0004
	COMPLEX CJI,ET1,ET2,EP1,EP2,EPPS,ETTS,EPTS,ETPS,ZS,VP,VT	0005
	COMPLEX C(1),CJ(1),EP(1),ET(1),EPP(1),ETT(1),ZLD(1)	0006
	COMPLEX ETA,GAM,CGD(1),SGD(1),CG(1)	0007
	DIMENSION IA(1),IB(1),I1(1),I2(1),I3(1),ND(1),MD(INM,4)	0008
	DIMENSION D(1),X(1),Y(1),Z(1)	0009
	DATA PI,TP/3.14159,6.28318/	0010
	CJI=-4.*PI/(ETA*GAM)	0011
	GGG=REAL(1./ETA)	0012
	THR=.0174533*TH	0013
	CTH=COS(THR)	0014
	STH=SIN(THR)	0015
	PHR=.0174533*PH	0016
	CPH=COS(PHR)	0017
	SPH=SIN(PHR)	0018
	DO 130 I=1,N	0019
	ETT(I)=(.0,.0)	0020
130	EPP(I)=(.0,.0)	0021
	DO 140 K=1,NM	0022
	KA=IA(K)	0023
	KB=IB(K)	0024
	CALL GFF(X(KA),Y(KA),Z(KA),X(KB),Y(KB),Z(KB),D(K)	0025
	2,CGD(K),SGD(K),CTH,STH,CPH,SPH,GAM,ETA,ET1,ET2,EP1,EP2)	0026
	NDK=ND(K)	0027
	DO 140 II=1,NDK	0028
	I=MD(K,II)	0029
	FI=1.	0030
	IF(KB.EQ.I2(II))GO TO 136	0031
	IF(KB.EQ.I1(II))FI=-1.	0032
	EPP(I)=EPP(I)+FI*EP1	0033
	ETT(I)=ETT(I)+FI*ET1	0034
	GO TO 140	0035
136	IF(KA.EQ.I3(II))FI=-1.	0036
	EPP(I)=EPP(I)+FI*EP2	0037
	ETT(I)=ETT(I)+FI*ET2	0038
140	CONTINUE	0039
	EPPS=(.0,.0)	0040
	ETTS=(.0,.0)	0041
	IF(INC.EQ.0)GO TO 200	0042
	IF(INC.EQ.2)GO TO 170	0043
	DO 150 I=1,N	0044
	ET(I)=ETT(I)*CJI	0045
150	EP(I)=EPP(I)*CJI	0046
	CALL SQROT(C,EP,0,I12,N)	0047
	I12=2	0048
	CALL SQROT(C,ET,0,I12,N)	0049
	CALL RITE(IA,IB,INM,IWR,I1,I2,I3,MD,ND,NM,EP,CG)	0050
	CALL GDISS(AM,CG,CMM,D,PDIS,GAM,NM,SGD,ZLD,ZS)	0051
	CALL RITE(IA,IB,INM,IWR,I1,I2,I3,MD,ND,NM,ET,CG)	0052
	CALL GDISS(AM,CG,CMM,D,TDIS,GAM,NM,SGD,ZLD,ZS)	0053
	ACSP=PDIS/GGG	0054
	ACST=TDIS/GGG	0055
	PIN=.0	0056
	TIN=.0	0057
	DO 164 I=1,N	0058
	VP=CJI*EPP(I)	0059
	VT=CJI*ETT(I)	0060
	PIN=PIN+REAL(VP*CONJG(EP(I)))	0061
164	TIN=TIN+REAL(VT*CONJG(ET(I)))	0062

Fig. 21a. Subroutine GFFLD

	ECSP=PIN/GGG	0063
	ECST=TIN/GGG	0064
	SCSP=ECSP-ACSP	0065
	SCST=ECST-ACST	0066
170	EPTS=(.0,.0)	0067
	ETPS=(.0,.0)	0068
	DO 180 I=1,N	0069
	EPPS=EPPS+EP(I)*EPP(I)	0070
	EPTS=EPTS+EP(I)*ETT(I)	0071
	ETTS=ETTS+ET(I)*ETT(I)	0072
180	ETPS=ETPS+ET(I)*EPP(I)	0073
	SPPM=2.*TP*(CABS(EPPS)**2)	0074
	SPTM=2.*TP*(CABS(EPTS)**2)	0075
	STPM=2.*TP*(CABS(ETPS)**2)	0076
	STTM=2.*TP*(CABS(ETTS)**2)	0077
	RETURN	0078
200	DO 260 I=1,N	0079
	ETTS=ETTS+CJ(I)*ETT(I)	0080
260	EPPS=EPPS+CJ(I)*EPP(I)	0081
	APP=CABS(EPPS)	0082
	ATT=CABS(ETTS)	0083
	GPP=4.*PI*APP*APP*GGG/GG	0084
	GTT=4.*PI*ATT*ATT*GGG/GG	0085
	RETURN	0086
	END	0087

Fig. 21b. Subroutine GFFLD

previous call for the backscattering situation with INC = 1. Thus, a bistatic call must be preceded by a backscatter call.

EPP(I) and ETT(I) denote the phi-polarized and theta-polarized far-zone fields of dipole mode I with unit terminal current. In a backscattering situation, the excitation voltages EP(I) and ET(I) are obtained by multiplying EPP and ETT by the constant CJI. (See Eqs. 38, 39 and 40 in Reference 1.) Then calls are made to SQROT which stores the solution (the induced loop currents) in EP(I) and ET(I). RITE is called for the branch currents CG(J), and GDISS is called for the time-average power dissipated in the imperfectly conducting wire and the lumped loads. This power is denoted PDIS and TDIS for phi-polarized and theta-polarized incident waves, respectively.

In scattering problems, the incident plane wave has unit electric field intensity at the coordinate origin. GGG denotes the time-average power density of the incident wave at the origin. ACSP and ACST denote the absorption cross sections for the phi and theta polarizations.

PIN and TIN denote the time-average power input to the wire structure, delivered by the equivalent voltage generators VP and VT at the terminals. PIN and TIN apply for the phi and theta polarizations, respectively. The time-average power input is regarded as the sum of the time-average power dissipated (in the wire and the lumped loads) and the time-average power radiated or scattered by the wire. ECSP and ECST denote the extinction cross sections and SCSP and SCST are the scattering cross sections.

The distant field is calculated in the DO LOOP ending with statement 180 for scattering situations, and in the DO LOOP ending with statement 260 for the antenna situation. In these fields, the range dependence is suppressed as in Eq. (1).

The radar cross sections (echo areas) SPPM, SPTM, STPM and STTM are defined as in Eq. 72 of Reference 1 with the incident power density (S_i or GGG) evaluated at the coordinate origin. The user selects the location of the origin when supplying the input data for the coordinates of all the points on the wire.

For an antenna, the following definition is employed for the power gain:

$$(6) \quad G_p(\theta, \phi) = \lim_{r \rightarrow \infty} 4\pi r^2 e^{2\alpha r} S(r, \theta, \phi) / P_i$$

where P_i (or GG in the program) denotes the time-average power input and $S(r, \theta, \phi)$ is the time-average power density in the radiated field. For an antenna in a lossless medium, α vanishes and Eq. (6) reduces to the standard definition of power gain. Without the factor $e^{2\alpha r}$ in Eq. (6), the power gain would vanish for a finite antenna in a conducting medium. GPP and GTT denote the power gains associated with the phi-polarized and theta-polarized components of the field, respectively.

APPENDIX 15. Subroutine GFF

Subroutine GFF, listed in Fig. 22, uses the equations in Appendix 2 of Reference 1 to calculate the far-zone field of a sinusoidal electric monopole. The monopole has endpoints (X_A, Y_A, Z_A) and (X_B, Y_B, Z_B) . EP1 and ET1 denote E_ϕ and E_θ for the mode with unit current at (X_A, Y_A, Z_A) . EP2 and ET2 denote the fields for the mode with unit current at (X_B, Y_B, Z_B) . The range dependence is suppressed as in Eq. (1). The far field vanishes in the endfire direction where $GK = 0$.

SUBROUTINE GFF(XA,YA,ZA,XB,YB,ZB,D,	0001
2CGD,SGD,CTH,STH,CPH,SPH,	0002
2GAM,ETA,ET1,ET2,EP1,EP2)	0003
COMPLEX ET1,ET2,EP1,EP2,GAM,ETA	0004
COMPLEX GO,CGD,SGD,EGD	0005
COMPLEX EGFA,EGFB,EGGD,ESA,ESB	0006
COMPLEX CST	0007
FP=12.56637	0008
XAB=XB-XA	0009
YAB=YB-YA	0010
ZAB=ZB-ZA	0011
CA=XAB/D	0012
CB=YAB/D	0013
CG=ZAB/D	0014
G=(CA*CPH+CB*SPH)*STH+CG*CTH	0015
GK=1.-G*G	0016
ET1=(.0,.0)	0017
ET2=(.0,.0)	0018
EP1=(.0,.0)	0019
EP2=(.0,.0)	0020
IF(GK.LT..001)GO TO 200	0021
FA=(XA*CPH+YA*SPH)*STH+ZA*CTH	0022
FB=(XB*CPH+YB*SPH)*STH+ZB*CTH	0023
EGFA=CEXP(GAM*FA)	0024
EGFB=CEXP(GAM*FB)	0025
EGGD=CEXP(GAM*G*D)	0026
CST=ETA/(GK*SGD*FP)	0027
ESA=CST*EGFA*(EGGD-G*SGD-CGD)	0028
ESB=CST*EGFB*(1./EGGD+G*SGD-CGD)	0029
T=(CA*CPH+CB*SPH)*CTH-CG*STH	0030
P=-CA*SPH+CB*CPH	0031
ET1=T*ESA	0032
ET2=T*ESB	0033
EP1=P*ESA	0034
EP2=P*ESB	0035
200 CONTINUE	0036
RETURN	0037
END	0038

Fig. 22. Subroutine GFF